Lucas Matana Luza

A CONTRIBUTION TO THE IN-ORBIT VALIDATION OF A RADIATION-HARDENED COMMUNICATION PLATFORM TO BE USED IN SMALL SATELLITES

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To my family, especially my parents who give me support for the fulfillment of my dreams.

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"Forget your lust for the rich man's gold All that you need is in your soul And you can do this, if you try All that I want for you, my son, is to be satisfied".

Gary Rossington / Ron Van Zant

RESUMO

Durante a última década, o número de missões espaciais utilizando nanossatélites aumentou consideravelmente devido a utilização do padrão CubeSat. O padrão CubeSat trouxe uma opção viável tanto para a comunidade acadêmica quanto para as empresas, uma vez que traz um orcamento consideravelmente reduzido guando comparado a modelos desenvolvidos por grandes corporações. Entretanto, diferentes protocolos de comunicações são empregados, diminuindo a integração entre as missões espaciais. Contudo, os protocolos definidos pelo CCSDS surgem como uma forte alternativa para quebrar essa barreira de integração, uma vez que já foram utilizados em mais de 900 missões espaciais. Com isso, este trabalho descreve o desenvolvimento de um módulo de comunicação para nanossatélites seguindo as recomendações propostas pelo CCSDS. Este trabalho tem como dispositivo alvo um novo FPGA endurecido contra os efeitos da radiação (radiation-hardened). Esse FPGA foi desenvolvido por uma companhia francesa e tem um grande potencial de utilização em novas missões espaciais devido ao fato de não ter as restrições provindas das regulamentações ITAR e EAR. Para alcancar os objetivos, o trabalho faz uso de uma unidade de telemetria e telecomandos e desenvolve uma arquitetura para um OBDH com o propósito de validar os telecomandos recebidos e gerar telemetria, tanto no aspecto de dados de status quanto para dados científicos. A implementação proposta por este trabalho tem como base a arquitetura do Pavload-X, que foi desenvolvido para ser operado como carga útil juntamente com o nanossatélite FloripaSat-I. Testes em hardware-in-the-loop aplicados a casos de usos do sistema mostraram a validade dos fluxos de telecomando e telemetria, sendo também verificadas a alocação de recursos do FPGA pela implementação e a utilização do Static Time Analisus para definir a viabilidade da aplicação.

Palavras-chave: CCSDS; Nanossatélite; Módulo de Comunicação; FPGA resistente a radiação.

RESUMO EXPANDIDO

Introdução

Nos últimos anos, o Grupo de Sistemas Embarcados (GSE) da UFSC vem realizando atividades de pesquisa e desenvolvimento visando a concepcão de uma missão completa de um CubeSat. O primeiro satélite desenvolvido no âmbito dessa iniciativa recebeu a denominação FloripaSat-I. Levando em consideração os trabalhos do grupo, pode-se observar que FPGAs são comumente utilizados no desenvolvimento de componentes imprescindíveis nas missões, por exemplo, os computadores de bordo e unidades de comunicação. Entretanto, técnicas de tolerância à radiação são necessárias uma vez que prótons e elétrons presos nos cinturões de radiação da Terra, tal qual os raios cósmicos, representam um dano real para a eletrônica, que deve operar de maneira confiável no ambiente espacial. Nesse ponto, o FPGA tolerante à radiação desenvolvido pela empresa francesa NanoXplore, apresenta a confiabilidade necessária para este tipo de missão e ultrapassa as barreiras impostas pelas regulações do ITAR e EAR que trazem dificuldades nas aquisições de componentes para missões espaciais providos por empresas estadunidenses. Além do mais, a padronização dos protocolos de comunicação para nanosatélites tem o objetivo de remover barreiras entre missõs espaciais, neste aspecto, as recomendações propostas pelo CCSDS (do inglês The Consultative Committee for Space Data Systems) trazem uma padronização para comunicações espaciais, sendo que as mesmas já foram utilizadas em mais de 900 missões.

Objetivos

O trabalho tem como objetivo geral o desenvolvimento de um módulo de comunicação para CubeSat utilizando as recomendações CCSDS aplicado em um FPGA tolerante à radiação. Os objetivos específicos deste trabalho contemplam: a utilização de uma unidade de telemetria e telecomandos e o desenvolvimento do componente de manipulação de dados (OBDH) compatível com as recomendações CCSDS para validar os telecommandos e gerar telemetria; E a aplicação desta unidade em um estudo de caso baseado na arquitutera do Payload-X, que foi desenvolvido para ser uma carga útil na missão FloripaSat-I.

Metodologia

Primeiramente, esse trabalho implementa o módulo comunicação uti-

lizando a linguagem de descrição de hardware VHDL, sendo aplicadas máquinas de estados para controle dos blocos lógicos desenvolvidos. A codificação das máquinas de estado segue o modelo proposto pela Xilinx no XTS User Guide e a implementação é fisicamente testada com o auxílio da placa de desenvolvimento da empresa NanoXplore. Foi possível testar o sistema utilizando a ferramenta Cortex que é amplamente utilizada na área espacial para sistemas de comunicação. O módulo de comunicação é composto pela unidade de telemetria e telecomandos (UTMC) e de um manipulador de dados (OBDH). A UTMC utilizada é fruto das pesquisas e desenvolvimento do grupo GSE e, como contribuição desse trabalho de mestrado, recebeu alterações para que pudesse ser utilizada em conjunto com o BRAVE FPGA, com o foco em modificações de processos para atingir a frequência de operação desejada e reformulação da descrição para evitar a inferência de latches na aplicação. A UTMC é responsável pela codificação e decodificação dos dados e realiza a manipulação dos dados na camada de Data Link proposta pelo CCSDS. O OBDH proposto neste trabalho tem como objetivo a validação dos telecomandos recebidos, e a geração de telemetria com dados de status e científicos. Para isso a arquitetura possui diferentes camadas que realizam processos de verificação, codificação, decodificação e empacotamento dos dados. O módulo de comunicação foi aplicado em um estudo de caso utilizando a plataforma denominada Pavload-X, que foi desenvolvida durante esse trabalho de mestrado em conjunto com outros mestrados realizados por colegas do grupo de pesquisa. O Payload-X tem como uma de suas características permitir a modificação da configuração presente no FPGA por meio de um mecanismo de reconfiguração que utiliza dados recebidos durante o voo. O Pavload-X será integrado a missão FloripaSat na forma de carga útil. E é previsto a utilização de dois diferentes modos de operação focando a validação da estrutura desenvolvida.

Resultados e Discussão

Este trabalho apresentou o desenvolvimento de um módulo de comunicação para pequenos satélites baseado em tecnologia de hardware reconfigurável (FPGA). O foco principal deste trabalho foi a construção de uma arquitetura baseada nas recomendações do CCSDS. A arquitetura proposta visa um equilíbrio para atender às demandas específicas do estudo de caso e ser genérica o suficiente para ser aplicada em outros cenários. Diante disso, todos os blocos OBDH foram construídos para integrar este modelo com novos blocos que atenderão às demandas de uma missão futura. Além disso, a implementação foi simulada principalmente usando a ferramenta ModelSim, e foi testada fisicamente com a ajuda da placa de desenvolvimento da empresa NanoXplore, fabricante do FPGA utilizado. Como os resultados apresentados são satisfatórios para a aplicação proposta, é possível concluir que o trabalho alcançou os objetivos pretendidos. É importante ressaltar que o trabalho tem continuidade no enfrentamento da missão FloripaSat-I, uma vez que está embutido na arquitetura desenvolvida para o Pavload-X, que é o resultado do trabalho conjunto com os colegas do grupo de pesquisa. Durante o desenvolvimento do trabalho foi possível deteminar a possibilidade de trabalhos futuros: como a validação da arquitetura fazendo uso da placa desenvolvida para o Payload-X que tem previsão de ficar pronta nos dias que precedem a finalização deste trabalho; a implementação do algoritmo de controle e teste das memórias SRAM; aplicação dos serviços propostos pelo padrão pelo ECSS PUS; e o estudo e implementação de um barramento de comunicação On-Chip, como por exemplo, o padrão AMBA desenvolvido pela ARM.

Palavras-chave: CCSDS; Nanossatélite; Módulo de Comunicação; FPGA resistente a radiação.

ABSTRACT

During the last decade, the number of space missions using nanosatellites has increased considerably, this is due to the use of the Cube-Sat standard that has brought a viable option for both the academic community and the companies since it brings a considerably reduced budget when compared to models developed by large corporations. However, different communication protocols are employed, reducing the integration between these space missions. Though, the protocols defined by the CCSDS appear as a strong alternative to break this integration barrier since they have already been used in more than 900 space missions. As a result, this work describes the development of a communication module for nanosatellites following the recommendations proposed by the CCSDS. This work has as its target device a new radiation-hardened FPGA developed by a French company that has a high potential usage in new space missions since it does not have the restrictions coming from the ITAR and EAR regulations. The work makes use of a telemetry unit and remote controls and develops a platform for an OBDH to validate the incoming telecommands and to generate telemetry, both in terms of status data and scientific data. The implementation proposed by this work is based on the Payload-X architecture, which was developed to be used as a payload in the FloripaSat-I nanosatellite. Hardware-in-the-loop tests applied to system use cases showed the validity of telecommand and telemetry flows, as well as the allocation of FPGA resources through implementation and the use of Static Time Analysis to define the feasibility of the application.

Keywords: CCSDS; Nanosatellite; Communication Module; Radiation-Hardened FPGA.

LIST OF FIGURES

Figure 1	Van Allen belts with respect to the flux and Earth ra-	
dius. In th	he left is shown the proton contribution and the right, the	
electron co	ontribution.	36
Figure 2	A typical FPGA internal components	40
Figure 3	An FPGA basic logic element	40
Figure 4	Overview of the Island-style interconnection scheme. \ldots	41
Figure 5	Hierarchical routing architecture	42
Figure 6	A Radiation-Hardened By Design latch.	43
Figure 7	Full Time Redundancy scheme	44
Figure 8	Full Hardware Redundancy, a TMR One-Bit Counter	45
Figure 9	Resistivity hardened CMOS SRAM cell design	45
Figure 10	DICE Cell.	46
Figure 11	CubeSat classification according their volume	47
Figure 12	Unnumbered and Supervisory frame construction	49
Figure 13	Infomation frame construction	49
Figure 14	Structure of a NGHam frame	50
Figure 15	CCSDS protocols defined by layers	51
Figure 16	Space Packet Protocol structure	51
Figure 17	Space Packet Protocol primary header structure	52
Figure 18	TC Transfer Frame: (a) structural components; (b) pri-	
mary head	ler structure	53
Figure 19	TM Transfer Frame: (a) structural components; (b) pri-	
mary head	ler structure	54
Figure 20	COP representation constituting of variables, frame and	
report val	ues	54
Figure 21	CLTU data unit with the structure of a BCH codeword.	55
Figure 22	CLTU data unit with the structure of a LDPC codeword.	56
Figure 23	Space Packet Protocol secondary header structure, (a)	
telecomma	and and (b) telemetry	58
Figure 24	Space Packet Protocol user data structure	58
Figure 25	UTMC top level diagram.	59
Figure 26	Convolutional Encoder	61
Figure 27	Finite state machine with three processes diagram	68

Figure 28	NanoXplore Brave development kit	68
Figure 29	Cortex setup	69
Figure 30	OBDH top level diagram	70
Figure 31	Packetization layer finite state machine	72
Figure 32	Verification layer finite state machine	73
Figure 33	Route layer finite state machine	74
Figure 34	Configuration layer finite state machine	75
Figure 35	Status layer finite state machine	76
Figure 36	Report layer finite state machine	77
Figure 37	Transfer layer finite state machine	78
Figure 38	Memory control finite state machine	79
Figure 39	I^2C controller finite state machine	80
Figure 40	Test setup	81
Figure 41	Payload-X interconnection diagram.	82
Figure 42	Contribution for the Payload-X project	83
Figure 43	Payload-X board	84
Figure 44	Nominal operation mode top level diagram	86
Figure 45	Advanced operation mode top level diagram	86
Figure 46	Use case CCSDS telecommand diagram	87
Figure 47	Use case CCSDS telemetry diagram	88
Figure 48	Use case bitstream upload diagram	89
Figure 49	Use case bitstream status request diagram	91
Figure 50	Use case bitstream status reply diagram	92
Figure 51	Use case bitstream swap version diagram	93
Figure 52	Data and clock delay between a source and destination	
register		96
Figure 53	OBDH register configuration telecommand structure	97
Figure 54	Results captured from the VHDL simulation of the OBDH $$	
register co	nfiguration flow (ModelSim)	98
Figure 55	OBDH register configuration flow: telecommand to be	0.0
sent		99
Figure 56	UBDH register configuration flow: expected telemetry	00
Figure 57	ORDH register configuration flow: expected telemetry	100
(second p	art) 1	01
veren he		

Figure 58 OBDH status request telecommand structure102
Figure 59 Results captured from the VHDL simulation of the OBDH $$
status request flow (ModelSim)103
Figure 60 OBDH request status flow: telecommand to be sent 104
Figure 61 OBDH request status flow: expected telemetry (first
part)
Figure 62 OBDH request status flow: expected telemetry (second
part)
Figure 63 Results captured from the VHDL simulation of the CPDU
procedure (ModelSim)108
Figure 64 CPDU telecommand structure
Figure 65 CPDU command flow: telecommand to be sent109
Figure 66 CPDU command flow: expected telemetry (first part). $.110$
Figure 67 CPDU command flow: expected telemetry (second part).111
Figure 68 Output 13ms pulse from a CPDU command112
Figure 69 CPDU telecommand structure with inseted error114
Figure 70 CPDU with error command flow: telecommand to be
sent
Figure 71 CPDU with error command flow: expected telemetry
(first part)116
Figure 72 CPDU with error command flow: expected telemetry
(second part)
Figure 73 OBDH command structure with wrong destination. $\ldots . 118$
Figure 74 Command flow of a OBDH command with wrong desti-
nation: telecommand to be sent 119
Figure 75 Command flow of a OBDH command with wrong desti-
nation: expected telemetry (first part)120
Figure 76 Command flow of a OBDH command with wrong desti-
nation: expected telemetry (second part)121
Figure 77 Packet received in CORTEX122

LIST OF TABLES

Table 1	Satelites classification according their mass	48
Table 2	CADU with different coding schemes	56
Table 3	Characterization of the systems presented in the related	
works		65
Table 4	CCSDS Telecommand step-by-step	88
Table 5	CCSDS Telemetry step-by-step.	89
Table 6	Bitstream upload step-by-step	90
Table 7	Bitstream status request step-by-step	91
Table 8	Bitstream status reply step-by-step	92
Table 9	Bitstream swap version step-by-step	93
Table 10	BCH cases according the <i>status</i> and <i>aff</i> signals1	.07
Table 11	Direct telecommand NACK error codes 1	13
Table 12	OBDH Verification NACK error codes 1	13
Table 13	Cortex report of the un-coded telemetry frame. $\ldots \ldots 1$	23
Table 14	Cortex report of the Reed-Solomon telemetry frame1	24
Table 15	Synthesis result: resource usage1	25
Table 16	Static Time Analisys results 1	26

LIST OF ABREVIATIONS AND SYMBOLS

ACG Automatic Control Gain ARM Advanced **RISC** Machine BCH Bose-Chaudhuri-Hocquenghem Big Re-programmable Array for Versatile Environments BRAVE CAD Computer Aided Design CADU Channel Access Data Unit CCSDS Consultive Committee for Space Data Systems CDMS Command and Data Management System CDS **CCSDS** Day Segmented CGRS Cosmic Galactic Rays CLBs Configurable Logic Blocks CLCW Communications Link Control Word CLTU Comunications Link Transmission Unit cm^3 Cubic centimeter CME **Coronal Mass Ejection** CMIC **Configuration Memory Integrity Check** COP **Communications Operation Procedure** CPDU Command Pulse Distribution Unit CRC Cyclic Redundancy Check CUC **CCSDS** Unsegmented DD Displacement Damage DMR. Dual Modular Redundancy DPRAM Dual-Port RAM DUT Device Under Test EDAC Error Detection and Correction EPS Electrical Power System ESA European Space Agency eVElectronvolt FARM Frame Acceptance and Reporting Mechanism FEC Forward Error Correction FOP Frame Operational Procedure FPGA Field Programmable Gate Arrays GeV Giga Electronvolt GEO Geostationary Orbit HDL Hardware Description Language HEO High Earth Orbit HUMAN Housekeeper and Update MANager

IP	Intellectual Property
keV	Kilo Electronvolt
km	Kilometer
LDPC	Low-Density Parity-Check
LEO	Low Earth Orbit
LFSR	Linear Feedback Shift Register
LUT	Look-up Table
MBU	Multiple-Bit Upset
MCU	Multiple-Cell Upset
MEO	Medium Earth Orbit
MeV	Mega Electronvolt
MOS	Metal Oxide Semiconductor
NGHam	Next Generation Ham
OBC	On-Board Computer
OBDH	On-Board Data Handling
I/O	Input/Output
PUS	Packet Utilization Standard
RHBD	Radiation-Hardened By Design
RS	Reed-Solomon
RTL	Register-Transfer Level
SAA	South Atlantic Anomaly
SBU	Single-Bit Upset
SEB	Single Event Burnout
SEEs	Single Event Effects
SEFI	Single Event Functional Interruption
SEGR	Single Event Gate Rupture
SEL	Single Event Latch-up
SET	Single Event Transient
SRAM	Static Random Access Memory
SSI	Synchronous Serial Interface
TET-R	Total-Elapsed-Time Recorder
TID	Total Ionizing Dose
TMR	Triple Modular Redundancy
UART	Universal Asynchronous Receiver/Transmitter
UTMC	Unit of Telemetry and Telecommand
VHDL	VHSIC Hardware Description Language

CONTENTS

1 INTRODUCTION	31
1.1 OBJECTIVE	32
1.2 DOCUMENT ORGANIZATION	33
2 BACKGROUND	35
2.1 RADIATION	35
2.1.1 Space Radiation Environment	35
2.1.2 Radiation Effects	37
2.1.2.1 Cumulative Effects	37
2.1.2.2 Single Event Effects	37
2.2 RECONFIGURABLE HARDWARE	38
2.2.1 FPGA Architecture	39
2.2.2 FPGA Radiation Hardened Architecture	43
2.2.3 BRAVE FPGA	46
2.3 SMALL SATELLITES - CUBESATS	47
2.4 COMMUNICATION PROTOCOLS FOR SMALL SATEL-	
LITES	48
2.4.1 AX.25 Protocol	48
2.4.2 NGHam Protocol	49
2.4.3 CCSDS Recommendations	50
2.4.3.1 Space Packet Protocol	51
2.4.3.2 TM and TC Space Data Link Protocols	52
2.4.3.3 TM and TC Synchronization and Channel Coding	55
2.5 PACKET UTILIZATION STANDARD	56
2.6 UNITY OF TELEMETRY AND TELECOMMAND	59
3 RELATED WORKS	63
4 PROPOSED COMMUNICATION MODULE	67
4.1 MATERIALS AND METHODS	67
4.1.1 Hardware Description Language	67
4.1.2 Finite State Machine Coding	67
4.1.3 Development Kit	68
4.1.4 Cortex	69
4.2 ON-BOARD DATA HANDLING	69
4.2.1 Packetization Layer	72
4.2.2 Verification Layer	73
4.2.3 Route Layer	74
4.2.4 Configure Layer	75
4.2.5 Status Layer	76

4.2.6 Report Layer	77
4.2.7 Transfer Layer	78
4.2.8 Memory Control	79
4.2.9 I ² C Controller	80
4.3 TEST SETUP	81
4.4 PAYLOAD-X	82
4.5 FLORIPASAT-I INTEGRATION	84
4.5.1 Nominal Operation Mode	85
4.5.2 Advanced Operation Mode	85
4.5.3 Data Flow	87
4.5.3.1 CCSDS Telecommand	87
4.5.3.2 CCSDS Telemetry	87
4.5.3.3 Bitstream Upload	89
4.5.3.4 Bitstream Status Request	91
4.5.3.5 Bitstream Status Reply	92
4.5.3.6 Bitstream Swap Version	93
5 RESULTS	95
5.1 UTMC PORTING TO BRAVE FPGA	95
5.2 USE CASES	96
5.2.1 OBDH Register Configuration Flow	97
5.2.2 OBDH Status Request Flow	102
5.2.3 CPDU Command Flow	107
5.2.4 Negative-acknowledgement Error Codes	112
5.3 CORTEX RESULTS	122
5.4 SYNTHESIS RESULTS	125
6 CONCLUSION	127
6.1 ACADEMIC PRODUCTION	127
6.2 FUTURE WORKS	128
References	131
APPENDIX A – FSM VHDL Coding Example	141

1 INTRODUCTION

Since its set up in 2002, the activities performed at the the Embedded Systems Group (GSE) are in the context of research and development in the aerospace area. For the last 10 years, the research focus has been on CubeSat nano-satellites, which have a mass between 1 and 10 kg. These small sized satellites are considered a viable option for the academic community, as they require reduced budgets when comparing to the models developed by large corporations (RAZZAGHI, 2012). GSE currently works on a project called FloripaSat-I that aims the development of a complete space mission and proposes a CubeSat 1U with the following subsystems: the onboard computer; electrical power system; communications systems; pure passive attitude control; and payloads (VILLA et al., 2014; SLONGO et al., 2016).

Therefore, FPGAs with features of high performance, high density and hardened to radiation are in great demand. In this direction, NanoXplore SAS implemented a 65 nm complementary metal-oxidesemiconductor (CMOS) FPGA with radiation protection up to 100 000 rads and 60 $MeV \cdot cm^2 \cdot mg^{-1}$, multiple voltage level inputs and outputs, and double data rate type 2 (DDR2) memory support (NANOXPLORE, 2018a). Those devices are called Big Re-programmable Array for Versatile Environments (BRAVE) and because of their recent generation have never been tested in orbit.

This new radiation-hardened FPGA developed by NanoXplore can open new possibilities to new space missions since it does not imply in the regulations from the ITAR (International Traffic in Arms Regulations) and EAR (Export Administration Regulations). Then, it is valid to explicit that the ITAR and EAR are two important United States export control regulations which cover from military products to software and hardware, including space-related technology. These regulations may involve difficult access to components such as radiationhardened FPGAs manufactured in the USA due to the bureaucratic procedures (COOK, 2010).

Furthermore, the standardization of communications protocol for nanosatellites has the aim to remove integration barriers between space missions. On this aspect, the international recommendations of the Consultative Committee for Space Data Systems (CCSDS) bring a standardization for space communications. These recommendations have already been used in more than 900 space missions (CCSDS, 2019).

The detection and correction of errors aim at bringing reliability

and integrity to the telemetry and remote control system of a space mission. Different error detection and correction codes (EDACs) can be applied in the frame structures that are used with the communications modules such as the BCH (Bose–Chaudhuri–Hocquenghem), Reed-Solomon, Convolutional, CRC (Cyclic Redundancy Check) and so forth (MOON, 2005).

1.1 OBJECTIVE

The global objective of this work is to perform the research and development of a communication module for CubeSats using the CCSDS protocols, targeting their use in a radiation-tolerant FPGA.

Also, this work proposes the utilization of a Unit of Telemetry and Telecommand (UTMC) previously developed to the Brazilian Institute of Space Research (INPE), in the context of a partnership between GSE and the companies Innalogics and AEL Systems. An important contribution of the developed research, was the adaptations performed in the UTMC source code, in order to port it to the new radiationhardened FPGA considering the necessary modifications to achieve the expected performance. Also, this work should provide the connection between UTMC and an OBDH with reduced functionality, aiming the validation of the received Telecommands and the generation of telemetry data from sensors by using an I²C communication interface.

A study case was developed targeting the integration of the communication module with the remaining FloripaSat boards, as a payload. To allow this case study, this work targets the development of the FPGA functionality provided by the Payload-X architecture.

The specific objectives to achieve the general goal include:

- UTMC porting to the BRAVE FPGA;
- Development of an OBDH based on the ECSS PUS recommendations;
- Integration between the UTCM and the OBDH;
- Application of the communication module in a case study based on the Payload-X platform;
- Development of a test setup for the application;
- Simulations and physical test to validate the communication module;

1.2 DOCUMENT ORGANIZATION

Chapter 2 gives an overview of the space radiation environment and the types of effects that are caused on electronic devices. The chapter also includes topics about reconfigurable hardware, communications protocols in small satellites and the CCSDS applications in this scope. Next, Chapter 3 presents related works that focus on different communications protocols applied in CubeSat's missions.

Chapter 4 describes the development process to create the application ported into the BRAVE FPGA and its simulation framework. It also describes how the integration with the FloripaSat-I mission and its use cases is performed. After, in Chapter 5, the results are introduced and discussed. Finally, Chapter 6 presents the conclusions and suggestions for future works.

2 BACKGROUND

As the on-board systems of CubeSats are sensitive to radiation effects, in this chapter there is an overview of radiation sources and their effects on electronic devices. Since the work involves the use of an FPGA as the main application device, its architecture will be presented. Next, a section is dedicated to the CubeSat standard and the nano-satellites classifications. This section also presents a review of the communication protocols used in small satellites focusing on their structures and capabilities.

2.1 RADIATION

When exposed to a radiation space environment, the effects caused on electronic devices are one of the main concern for all type of satellites missions, from the Low Earth Orbit (LEO) to deep space (VELAZCO; FOUILLAT; REIS, 2007; NICOLAIDIS, 2011; FLEETWOOD; WINOKUR; DODD, 2000).

2.1.1 Space Radiation Environment

The space environment is classified into four categories accordingly with the origins: radiation belts, solar flares, solar wind, and galactic cosmic rays (VELAZCO; FOUILLAT; REIS, 2007).

The Earth radiations belts, known as Van Allen Belts, are composed of trapped electrons and protons and are divided into two zones. The inner belts contain electrons with energy between 1 and 10 MeV and high-energy protons (>100 MeV). The outer belt is mainly composed of high-energy electrons (VELAZCO; FOUILLAT; REIS, 2007; DYER, 2002). Figure 1 presents the radiation belts fluxes with respect to the Earth radius.

There is a region located in the South Atlantic area, where a high flux of energetic protons is present. This anomaly is caused because of displacement between the magnetic and geographic axes. This region is known under the name of South Atlantic Anomaly (SAA) and a spacecraft is exposed to its radiation effects at altitudes about 1000 km (PETERSEN, 2011; DYER, 2002).

During the years close to the solar maximum, the solar flares

are a sporadic source of radiation and are defined by two types of events. The first one is the Coronal Mass Ejection (CME) that could remain for several days and emits high-energy protons with energies of hundreds MeV. The second type is the emission of heavy-ions with a energy range starting from several tens of MeV up to hundreds GeV per nucleon (VELAZCO; FOUILLAT; REIS, 2007; DYER, 2002; PETERSEN, 2011).

The solar wind particles have not significant energy when compared with solar flares and in the Geostationary Orbit (GEO) leads to electrons with energies around 2 keV and ions at 10 keV, while in the LEO the energies are even lower and incapable to inducing significant charge deposition (VELAZCO; FOUILLAT; REIS, 2007; PETERSEN, 2011).

Cosmic Galactic Rays (CGRs) are composed of high energetic heavy-ions (1%), protons (83%), helium nuclei (13%) and electrons (3%). The ions' energy is very high, with the most energetic ion ever detected having an energy of 3×10^{20} eV. Therefore, the primary concern in space applications is the effect of ionizing particles in the energy range of 1-20 GeV/nucleon (VELAZCO; FOUILLAT; REIS, 2007).

Figure 1 - Van Allen belts with respect to the flux and Earth radius. In the left is shown the proton contribution and the right, the electron contribution.



Source: (DYER, 2002).
2.1.2 Radiation Effects

The space radiation environment affects microelectronics devices with cumulative effects that are characterized by the Total Ionizing Dose (TID) and Displacement Damage (DD), but also by instantaneous and temporary effects defined as Single Event Effects (SEEs), which are due to the penetration of a single ionizing particle that causes issues in sensitive nodes of the circuits (PETERSEN, 2011; DYER, 2002).

2.1.2.1 Cumulative Effects

The TID is a long-term failure mechanism and is defined as the energy deposited per mass unit of material and that is generally expressed by the unit rad (in the SI) and that degrades the function of the device until its operation is definitely compromised (DYER, 2002; DOWD, 2003). The cumulative dose can cause shift in the transistor threshold voltage, increase of the leakage currents (power consumption), and also the degradation of the gain in bipolar devices (DYER, 2002).

According to Schwank (1994) and Johnston (2010), the DD effect is caused by the collision between high-energy protons and one atom generating a recoil of the same from its lattice site. When the transfer energy of the particle is high enough, the hitted atom can be freed from its interstitial site. The minimum energy required for this is called the displacement threshold energy. The accumulation of this effect can create a large fault cluster. Dyer (2002) cites examples of damage caused by this effect, as a reduction in bipolar transistor gain, reduction of the efficiency in solar cells, light emitting diodes and photodetectors, charge transfer inefficiency in charge coupled devices and resolution degradation in solid-state detectors.

2.1.2.2 Single Event Effects

A single ionizing particle (heavy ion, electrons, and protons) strike may induce a Single Event Effect, that occurs instantaneously and is due to a deposed charge that is collected by transistor electrodes of the device (NICOLAIDIS, 2011; VELAZCO; FOUILLAT; REIS, 2007; PE-TERSEN, 2011).

The SSEs are classified as soft-errors, since they cause data cor-

ruption that can be mitigated for example by a power cycle, a reset, or other erase operations. However, in some cases, they may lead to hard-errors that cause permanent device failure. The SSEs are divided into the following categories (NICOLAIDIS, 2011; VELAZCO; FOUILLAT; REIS, 2007; PETERSEN, 2011):

- Single-bit Upset (SBU) [Soft-error]: a transient event that causes a bit-flip (upset) in a memory cell or a latch;
- Multiple-Bit Upset (MBU) [Soft-error]: transient event caused by a single ionizing particle that generates two or more bit-flips in the same word;
- Multiple-Cell Upset (MCU) [Soft-error]: transient event caused by a single ionizing particle that generates two or more bit-flips in the cell array;
- Single Event Transient (SET) [Soft-error]: the collected free carriers generated by an ionizing particle may generate a current or voltage transient that may lead to an error, especially in combinational logic circuits;
- Single Event Functional Interruption (SEFI) [Soft-error]: this event is due a perturbation in control signals (e.g. registers, clock signals, reset signals) and cause a loss of functionality;
- Single Event Latch-up (SEL) [Hard/Soft-error]: when a particle hit triggers a parasitic thyristor that involves high-current consumption. This event requires a power cycle and may cause permanent damage due to thermal dissipation of large current;
- Single Event Gate Rupture (SEGR) [Hard-error]: this event occurs when a particle strike is capable of breaking the gate dielectric of a MOS transistor;
- Single Event Burnout (SEB) [Hard-error]: occurs when a strike causes a destructive burnout due to high-current and Joule effects.

2.2 RECONFIGURABLE HARDWARE

The reconfigurable hardware combines the flexibility of a software implementation with the hardware performance, which is potentially higher than a microcontroller solution, since the microcontroller must read each instruction from memory, interpret it and then execute the instruction (COMPTON; HAUCK, 2002). This is confirmed by Bezerra e Gough (2000) and Garcia et al. (2006), that show how the hardware-based implementations avoid the software overhead of fetch/decode/execute and use the resources to increase the parallelism.

The FPGA is an integrated circuit composed of an array of Configurable Logic Blocks (CLBs) and configurable interconnections that brings the reconfigurable hardware concept since their capability to be reprogrammed in a matter of microseconds. This reconfigurable logic blocks can implement combinational and sequential logic circuits. This implementation uses a Hardware Description Language (HDL) to define functionality in Register-Transfer Level (RTL) that will be synthesized into a netlist with the connectivity between the elements of the circuit (WAIN et al., 2006; FAROOQ; MARRAKCHI; MEHREZ, 2012).

2.2.1 FPGA Architecture

FPGA are devices composed of an array of circuitry's blocks that are generally comprised of three main components: configurable logic blocks (CLBs); configurable routing connections; I/O (inputs/outputs) blocks to provide external access to the chip. As presented in Figure 2, the logic blocks are distributed in an array format, the I/O blocks are arranged in the periphery of the grid, and the routing elements interconnect all these blocks (BEZERRA; LETTNIN, 2014; FAROOQ; MARRAKCHI; MEHREZ, 2012).

According to Farooq, Marrakchi e Mehrez (2012), the SRAM technology is the most used by commercial vendors since it allows easy re-programmability and use CMOS standard processes. For the SRAM-based FPGAs, the memory cells are used to store data in the logic blocks and to program the multiplexers that are used in the interconnections elements.

Most FPGAs use Look-up Tables (LUTs) in the CLB implementation. The LUTs implement any *n*-input truth table and are complemented with the use of a Flip-Flop, Figure 3 presents this generic structure in a SRAM-based FPGA (BEZERRA; LETTNIN, 2014; FAROOQ; MARRAKCHI; MEHREZ, 2012).

Wires and programmable switches composing the interconnections between the CLBs and I/Os and the basic FPGA architectures are defined by island-style and hierarchical interconnections schemes (FAROOQ; MARRAKCHI; MEHREZ, 2012).



Figure 2 – A typical FPGA internal components.

Source: based on (BEZERRA; LETTNIN, 2014; FAROOQ; MAR-RAKCHI; MEHREZ, 2012).

Figure 3 – An FPGA basic logic element.



Source: based on (BEZERRA; LETTNIN, 2014; FAROOQ; MAR-RAKCHI; MEHREZ, 2012).

Island-style organizes wiring segments that are connected through switches and connection blocks, however, as a drawback, this scheme uses between 80-90% of the total area of the chip, and logical blocks occupies only 10-20% (BETZ; ROSE; MARQUARDT, 1999). Figure 4 presents an architecture overview of the Island-style scheme, where the switches blocks are referred as 'SB', and the connection blocks 'CB' (FAROOQ; MARRAKCHI; MEHREZ, 2012).

Hierarchical routing architecture remains a tree-like structure, where the logic blocks are positioned as leaves in a tree, and there are several levels of interconnections. In the example presented in Figure 5, the architecture has four wire levels. This scheme uses two different kinds of switches (FAROOQ; MARRAKCHI; MEHREZ, 2012):

- Non-compressing: The number of tracks in the upper wire level is equal to the sum of the number of tracks in the lower wire level.
- Compressing: the number of tracks in the upper wire level is equal to the number of tracks in the lower wire level.



Figure 4 – Overview of the Island-style interconnection scheme.

Source: based on (FAROOQ; MARRAKCHI; MEHREZ, 2012).



Figure 5 – Hierarchical routing architecture.

Source: based on (FAROOQ; MARRAKCHI; MEHREZ, 2012).

2.2.2 FPGA Radiation Hardened Architecture

SRAM-based FPGAs are based on CMOS technology and they are susceptible to SEE (YUE et al., 2017). Several techniques can be introduced in these devices to increase their reliability and avoid the radiation effects. Besides the techniques used in the application upper level, Radiation-Hardened By Design (RHBD) devices mitigate these effects at hardware level (BATTEZZATI et al., 2009).

The work from Huang e Liang (2008) proposes an RHBD latch for sub-micron technologies that is immune to SEUs and SETs. According to the authors, the use of a dual interlocked scheme with internal feedback lines eliminate vulnerable nodes. Figure 6 present the architecture of the proposed latch.

Besides the technique for latch architectures presented by Huang e Liang (2008), we can found several different ways to face radiationinduced effects. Examples are presented in the works from Chen e Orailoglu (2007), Naseer e Draper (2006).





Source: based on (HUANG; LIANG, 2008).

Redundancy techniques can be divided into Full-Time Redundancy (that mitigates SET) and Full Hardware Redundancy (that mitigates SEU). The Full-Time Redundancy gets the output of combinatorial logic in three different moments that are defined by a small delay, and a majority voter (MAJ) is responsible for defining the correct output value (ZHANG; LIE, 2011; CALIN; NICOLAIDIS; VELAZCO, 1996). Figure 7 presents Full-Time Redundancy scheme.

Figure 7 – Full Time Redundancy scheme.



Source: adapted from (ZHANG; LIE, 2011).

According to Zhang e Lie (2011), Full Hardware Redundancy can be achieved by a Triple Modular Redundancy TMR scheme, where the logic is triplicated and a majority voter defines the correct output value. To further improve reliability, the scheme presented in Figure 8 implement the use of a triplicated voter and Flip Flops with feedback (ZHANG; LIE, 2011).

The application of EDACs is an efficient solution to protect memories against SEUs, with coding techniques applied to high capacity memory and periodically 'scrubbing' of the entire array (ZHANG; LIE, 2011; WANG et al., 1999; CALIN; NICOLAIDIS; VELAZCO, 1996).

At cell level different hardening techniques can be applied leading, for example, to the Hardened Memory Cells, Canaris Hardened Cells, DICE Hardened Memory Cell, and NASA Hardened Memory Cell (I and II). Focusing on commonly used techniques, the Hardened Memory Cells use two resistors between the inverters of an SRAM cell, and when a particle strikes one of the nodes these resistors will delay the spike propagation and prevent the bit-flip (ZHANG; LIE, 2011). Figure 9 presents this structure.



Figure 8 – Full Hardware Redundancy, a TMR One-Bit Counter.

Source: adapted from (ZHANG; LIE, 2011).

Figure 9 – Resistivity hardened CMOS SRAM cell design.



Source: adapted from (ZHANG; LIE, 2011).

According to Danilov et al. (2018), the DICE structure applies the idea that a charged particle has to achieve two or more sensitive nodes to generated a SEU. This principle was described by Calin, Nicolaidis e Velazco (1996), in which the authors refer that the SEU occurrence probability can be reduced if the sensitive node pairs are spaced on cell's layout.

Figure 10 – DICE Cell.



Source: (DANILOV et al., 2018).

2.2.3 BRAVE FPGA

This work leads to the use of the re-programmable European Radiation-Hardened FPGA developed by NanoXplore that is a fabless semiconductor company headquartered in France. This FPGA targets space applications and it is the result of a project between the Nano-Xplore (design, verify and FPGA tools) and the ST Microelectronics (foundry, IP provider and packing evaluation and qualification), involving the AirBus D&S and Thales Alenia as key alpha customers.

The NG-MEDIUM is the first FPGA of the Big Re-programmable Array for Versatile Environments (BRAVE) FPGA family and is radiation hardening by design in configuration memories and registers.

Register files and DPRAM (Dual-Port RAM) use EDAC to be able to detect and correct SEUs and detect MBUs. The configuration memory, user registers, SRAM and Flip-Flops use the DICE (Dual Interlocked Storage Cell) scheme that are more resilient to radiation than the conventional 6 transistor scheme. The clock tree applies DMR (Dual Modular Redundancy) strategy in the clock buffer and the distributed matrix system. Others logic cells apply TMR.

Besides this techniques, an embedded Configuration Memory Integrity Check (CMIC) scheme performs an automatic verification and repair in the configuration memory, this process is done at run time, and ECC protects the reference memory. As a result of the implemented techniques, the BRAVE FPGA has SEL immunity up to a LET of 60 MeV .cm²/mg, and can endure a TID up to 100 krads.

2.3 SMALL SATELLITES - CUBESATS

With the mission named OPAL (Orbiting Picosatelitte Automated Launcher), the launch of a small satellite developed by a student group at Stanford University created a new concept in the architecture of these devices. The mission was based on a microsatellite of 25 kg responsible for launching 6 picosatellites in the space (ENGBERG; OTA; SUCHMAN, 1995).

With the acquired experiences of the OPAL mission and the necessity to have picosatellites that have the requirements to a default launcher, the California Polytechnic State University and Stanford University started a new partnership project to create a standard picosatellite, the result of this project was the CubeSat² (HEIDT et al., 2000; PUIG-SUARI; TURNER; AHLGREN, 2001).

The CubeSat specification defines that a standard 1U is a cube of nearly $10 \times 10 \times 10 \times 10 \times 10^3$ with mass up to 1.33 kg and this unit could be combined forming a large spacecraft. CubeSats are classified according to their mass and volume and Figure 11 and Table 1 presents a concept of this classification (JOHNSTON, 2010; NASA CUBESAT LAUNCH INITIATIVE, 2017; THE CUBESAT PROGRAM, 2014).

Figure 11 – CubeSat classification according their volume.



Source: based on (JOHNSTON, 2010; NASA CUBESAT LAUNCH INITIATIVE, 2017; THE CUBESAT PROGRAM, 2014).

Classification	Mass (kg)
Pico	< 1
Nano	$1 \sim 10$
Micro	$10 \sim 100$
Mini	> 100

Table 1 – Satelites classification according their mass.

Source: adapted from (JOHNSTON, 2010).

According to Arnold, Nuzzaci e Gordon-Ross (2012) CubeSats usually are placed in Low Earth Orbit (LEO), which reach until 2.000 km from the surface. Orbits are classified related with their altitude, then, above LEO is Medium Earth Orbit (MEO) extended outward to 35.786 km, and right after is the High Earth Orbit (HEO) for altitudes higher the MEO limit.

Also, orbit patterns are related to their inclination (satellite orbit angle with respect to the equator) and the kind of orbit is directly related with the CubeSat power generation, since, for example, a Cube-Sat that orbits in a Sun-Synchronous orbit received a near-constant sun illumination and the power system is capable of providing a better power supply to the subsystems. Besides that, the size and weight restrictions of a CubeSat limits the use of batteries and external solar panels and thus the power available by the system (ARNOLD; NUZZACI; GORDON-ROSS, 2012).

2.4 COMMUNICATION PROTOCOLS FOR SMALL SATELLITES

2.4.1 AX.25 Protocol

The AX.25 protocol was developed by the radio amateur community since there was a need to define a protocol capable to accept and send data in a reliable way over a communication link between two terminals (BEECH; NIELSEN; TAYLOR, 1998). According to studies carried out in Muri e Mcnair (2012) and Klofas, Anderson e Leveque (2008) the AX.25 protocol has been chosen as principal choice for the small satellites' scopes.

The frame implemented in the AX.25 protocol is divided into three types: Information frame; Supervisory frame; and Unnumbered frame. The frames are composed of fields that identify the start and end of the frame ("Flag" field), the address of the source and destination ("Address" field), the frame type ("Control" field), the protocol identifier ("PID" field), the user data ("Info" field) and the frame check sequence that is calculated accordingly with ISO 3309 (HDLC) Recommendations and it is used to check if the received data is corrupted during the transmission ("FSC" field) (BEECH; NIELSEN; TAYLOR, 1998). Figures 12 and 13 present the two different constructions of the frame.

Figure 12 – Unnumbered and Supervisory frame construction.

First Bit Sent				
Flag	Address	Control	FCS	Flag
0111 1110	112/560 Bits	8 Bits	16 Bits	0111 1110

Source: (BEECH; NIELSEN; TAYLOR, 1998).

Figure 13 – Infomation frame construction.

First Bit Sent							
Flag	Address	Control	PID	Info.	FCS	Flag	
0111 1110	112/560 Bits	8 Bits	8 Bits	n*8 Bits	16 Bits	0111 1110	

Source: (BEECH; NIELSEN; TAYLOR, 1998).

This protocol was not conceived to embed telemetry and telecommands for space applications. Futher limitations are related to the fact that this protocol does not implement a Forward Error Correction (FEC) scheme, and it is unable of correcting errors in the received data (ZEIGER; SCHMIDT; SCHILLING, 2006).

2.4.2 NGHam Protocol

Next Generation Ham (NGHam) is a protocol that is based on the idea of the AX.25 protocol, however, improving the reliability using a FEC on the link layer. The FEC implements a Reed-Solomon (RS) algorithm with 16 or 32 check symbols (SKAGMO, 2014; KLEINSCHRODT et al., 2017; BIRKELAND; LØFALDLI, 2016).

Figure 14 presents the structure of a NGHam frame.

Preamble field is used to assure a timing requirement for proper interpretation since it is sending during the ramping up to the radio power amplifier and the receiver Automatic Control Gain (ACG) is



Figure 14 – Structure of a NGHam frame.

Source: (SKAGMO, 2014).

stabilizing. Also, a Sync Word is used for packet synchronization. The Size tag field identifies one of the seven possibilities of code block sizes. In the code block, the Header field is used to identify the padding size. Furthermore, the CRC field and the Parity data are introduced for error detection and correction implementations (SKAGMO, 2014).

2.4.3 CCSDS Recommendations

Historically, in the 1980s, CCSDS proposed an international standard for sending telemetry data using a variable-length data unit called Source Packet. This standard defined the continuous transmission of Source Packets generated from the instruments and subsystems of a spacecraft through Transfer Frames. Following this concept, another international standard was developed targeting the transmission of commands to a spacecraft with a data unit named TC Packet. In this case, the transmission is sporadic and uses a variable length Transfer Frame. Since the necessity to meet the requirements of the Advanced Orbiting Systems (e.g., the International Space Station) a third standard known as AOS was produced. The main concern in this standard is to add services for online transmission data (e.g., audio and video data) (CCSDS, 2014).

The above mentioned standards were restructured with the objective to define the protocols in a more structured way, generating: Space Packet Protocol; TM, TC, and AOS Space Data Link Protocols; and TM and TC Synchronization and Channel Coding.

The CCSDS defines several protocols, but this section only describes the recommendations that are within the scope of the work. These recommendations are distributed following a stack of protocols that are associated with layers. Figure 15 presents the protocols with their layer and their relationships.

Figure 15 – CCSDS protocols defined by layers.



Source: Adapted from (CCSDS, 2014).

2.4.3.1 Space Packet Protocol

The Space Packet Protocol defines a way of transferring application data over a network in a ground-to-space or space-to-ground link. This protocol is structured following the Figure 16 (CCSDS, 2003).

Figure 16 – Space Packet Protocol structure.



Source: Adapted from (CCSDS, 2003).

The Packet Primary Header is mandatory and consists of four fields, contiguously positioned. This header starts by defining which is the version of the packet that is related to the recommendation (e.g., CCSDS 133.0-B-1 defines version 1). The Packet Identification field specifies if the packet is telemetry or telecommand, and indicates the

presence or absence of the Packet Secondary Header; and the identification of the application process that has to receive or is sending this data. Packet Sequence Control defines if the packet is segmented and provides the sequential binary count of each Space Packet generated by the application process (mandatory for telemetry packets) or the packet name (that could be used in for telecommand packets. Packet Data Length contains a 16-bit count that represents the total number of bytes present in the Packet Data Field (CCSDS, 2003). Figure 17 presents its format.

Figure 17 – Space Packet Protocol primary header structure.

		Packet Identification		Packet Se		
Packet Version Pack Number Typ	Packet Type	Secondary Header Flag	Application Process Identifier	Sequence Flags	Packet Sequence Count or Packet Name	Packet Data Length
3 bits	1 bit	1 bit	11 bits	2 bits	14 bits	16 bits

Source: Adapted from (CCSDS, 2003).

2.4.3.2 TM and TC Space Data Link Protocols

Following the purpose to specify a link protocol for telecommand and telemetry to be used over ground-to-space or space-to-space communications, the CCSDS defines the CCSDS 132.0-B-2 TM Space Data Link Protocol (CCSDS, 2015a) and CCSDS 232.0-B-3 TC Space Data Link Protocol (CCSDS, 2015b), that defines the services, units and procedures of these protocols.

CCSDS (2015b) defines a protocol data unit named "TC Transfer Frame" that is structured as presented in Figure 18, while CCSDS (2015a) defines the "TM Transfer Frame", as shown in Figure 19. The TC and TM Space Data Link Protocol have as principal functions the segmentation and blocking of service data units and transmission control of service data units. This recommendation also provides an automatic retransmission mechanism procedure to ensure reliability on the received telecommands defined as Communications Operation Procedure (COP) that is specified in (CCSDS, 2010). The COP specification implements the ARQ-GBN¹. This mechanism to manage the retransmission, this procedure works as a closedloop. This loop is composed of the Frame Acceptance and Reporting Mechanism (FARM) (which is used in the receiver) and the Frame Operational Procedure (FOP) (which is applied in the sender). The FOP sends TC Frames to FARM, and the FARM provides a status report through the Communications Link Control Word (CLCW). This procedure uses sequence numbers 'N' within the frames, and variables 'V' that store these values. The Figure 20 presents this loop.

Figure 18 – TC Transfer Frame: (a) structural components; (b) primary header structure.



Number	Flag	Flag	Spare	Identifier	Identifier	Length	Number
2 bits	1 bit	1 bit	2 bits	10 bits	6 bits	10 bits	8 bits
∢		2 Bytes		·····>	∢ 2 B	ytes>	≺ -1 Byte →
			(b)			
a		. 1.0	/	004 51	`		

Source: Adapted from (CCSDS, 2015b).

¹ARQ-GBN (Automatic Repeat Query - Go-Back-N) is a retransmission mechanism used as a method of error control, and more information can be reached in the reference (Shu Lin; COSTELLO; MILLER, 1984)



Figure 19 – TM Transfer Frame: (a) structural components; (b) primary header structure.

Source: Adapted from (CCSDS, 2015b).

Figure 20 - COP representation constituting of variables, frame and report values.



Source: Adapted from (CCSDS, 2010)

In Figure 20, N(S) represents transmitted frame sequence number, N(R) is the next frame sequence number that has to be received, V(S) stores the sequence number, and V(R) stores the frame sequence number that has to be received. To validate the transmission the receptor has to satisfy: N(S) = V(R). 2.4.3.3 TM and TC Synchronization and Channel Coding

The recommendations CCSDS (2017a) and CCSDS (2017b) describes the synchronization and channel coding for telemetry and telecommand respectively and provide four functions for data transfer over a space link: error-control coding; synchronization; pseudo-randomizing (optional); and repeated transmissions (optional).

In the scope of the telecommand flow, the CCSDS recommendation 231.0-B-3 (CCSDS, 2017b) defines two error-control coding schemes, the first is the Bose–Chaudhuri–Hocquenghem (BCH) and the second is the Low-Density Parity-Check (LDPC), also, specifies these methods using a data unit called Communications Link Transmission Unit (CLTU) (CCSDS, 2017b).

The CLTU structure is composed of a start sequence, followed by the encoded data and a tail sequence. The encoded data consists of Transfer Frames according to CCSDS (2015b), and the tail sequence defines the end of the packet that is constructed to be a non-correctable pattern of the coding scheme. Figures 21 and 22 present the CLTU components when using, respectively, BCH and LDPC schemes.

Figure 21 – CLTU data unit with the structure of a BCH codeword.



Source: Adapted from (CCSDS, 2017b)

CCSDS (2017a) defines the use of four different coding schemes: Convolutional; Reed-Solomon; Turbo Coding; and LDPC. These algorithms methods are applied in the data unit called Channel Access Data



Figure 22 – CLTU data unit with the structure of a LDPC codeword.

Source: Adapted from (CCSDS, 2017b)

Unit (CADU). The Concatenated coding consists of a combination of a Reed-Solomon (outer code) with the Convolutional (inner code). Table 2 presents the CADU contents related to the coding scheme, and the containing data may or may not be randomized.

Table 2 – CADU with different coding schemes.

Applied Coding Scheme	CADU
Convolutional Coding	ASM and the Transfer Frame
Reed-Solomon Coding	ASM and Reed-Solomon codeblock
Concatenated Coding	ASM and Reed-Solomon codeblock
Turbo Coding	ASM and Turbo codeword
LDPC Coding	ASM and LDPC codeword

Source: Based on (CCSDS, 2017a).

2.5 PACKET UTILIZATION STANDARD

The Packet Utilization Standard (PUS) (ECSS-E-ST-70-41C) defines the application-level interface using the protocol established in the CCSDS Space Packet Protocol (CCSDS 133.0-B-1) targeting the utilization of telecommand and telemetry packets for remote monitoring and control of the subsystems (ECSS, 2016). For this, the standard defines a set of services with capabilities and structured data unit, which defines the content data of the Secondary Header and User Data fields presented in Figure 17.

The Secondary Header has two different formats, one is related to a telecommand, and the other is related to telemetry. Figure 23(a) presents the telecommand structure and the Figure 23(b) presents the telemetry one. When this secondary header is presented in the Space Packet, these fields shall follow, without a gap, the Packet Primary Header(ECSS, 2016).

The fields "PUS Version Number", "Message Type ID", and "Spare", are presented in both structures. The fields inform the applied ECSS version, identifies the type of the message that will define the structure of the user data field, and, when needed, the usage of the Spare field aiming at constraining the length to an integral number of words (ECSS, 2016).

Following the telecommand structure, "Ack Flags" are used to signalize which kind of report is required by the telecommand that is related with the service applied. These reports are defined by the Service 1 - Request Verification and possibilities four different types of report. "Source ID" field corresponds to the application process that hosts the "Service Type" (ECSS, 2016).

Related to telemetry structure, "Spacecraft Time Reference Status" defines a reference time report using CUC (CCSDS Unsegmented) or CDS (CCSDS Day Segmented) format, or, if it is not supported, has to be set as '0000'. A "Message Type Counter" related with the "Message Type" has to be maintained. "Time" reports the absolute time according to their reference, and the use of the "Spare" field is optional (ECSS, 2016).

User data field structure is composed of a variable field of source data that could be complemented by a spare field. Also, the "Packet Error Control" is optional and applies a CRC standard 16-bits checksum algorithm known as CRC-16-CCITT that used the polynomial generator $G(x) = x^{16} + x^{12} + x^5 + 1$ (ECSS, 2016). Figure 24 presents its structure.

Figure 23 – Space Packet Protocol secondary header structure, (a) telecommand and (b) telemetry.

∢	C Packet Secondary Header							
		Message	e Type ID					
TC PUS Version Number	Ack Flags	Service Type ID	Message Subtype ID	Source ID	Spare			
4 bits	4 bits	8 bits	8 bits	16 bits	fixed-size			

-- optional --

(a)

∢	TM Packet Secondary Header							
		Message	Type ID					
TM PUS Version Number	Spacecraft Time Reference Status	Service Type ID	Message Subtype ID	Message Type Counter	Destination ID	Time	Spare	
4 bits	4 bits	8 bits	8 bits	16 bits	16 bits	abs. time	fixed-size	

- -optional- -

(b)

Source: Adapted from (CCSDS, 2003) and (ECSS, 2016).

Figure 24 – Space Packet Protocol user data structure.

◀	User Data Field	····· >
Source Data	Spare	Packet Error Control
variable	fixed-size	16 bits

-----optional-----optional-----

Source: Adapted from (ECSS, 2016).

2.6 UNITY OF TELEMETRY AND TELECOMMAND

The GSE group developed in previous works a UTMC to be applied in the on-board computer of a Brazilian satellite. The main function of this implementation is to handle the TM and TC flow according to CCSDS recommendations and use the Data Link Layer presented in Section 2.4.3. The UTMC block is divided into two segments dedicated to telemetry data processing and telecommands handling. Figure 25 presents the UTMC architecture.





Source: adapted from (BEZERRA et al., 2010; BEZERRA; AZEVEDO; SILVA, 2011).

In the TC flow, when a TC is delivered to UTMC from the communication interface, the CLTU is decoding using BCH algorithms and performing the error detection and correction. When a Direct TC (DTC) is detected, the UTMC delivers it to a CPDU layer that interprets and generates a strictly controlled configurable time pulse

in one of the outputs available. Also, when a Routed TC (RTC) is detected, the UTMC delivers the Space Packet data to the OBDH system (BEZERRA et al., 2010; BEZERRA; AZEVEDO; SILVA, 2011).

The BCH decoder is implemented following the CCSDS recommendations which describes a BCH(63, 56) code using a generator matrix G(56, 63) and a parity matrix H(63, 7). The generator matrix G is composed by two submatrices, the identity matrix $I_1(56, 56)$ and the matrix $G_1(56, 7)$ that is composed by the syndrome values generated from the polynomial $g(x) = x^7 + x^6 + x^2 + 1$. The $H^T(63, 7)$ is formed by G_1 and I_1 matrices. For the decoding process, the identification of the syndrome value is given by the multiplication of the received message (BCH codeword) by the parity matrix H^T . A syndrome value equal zero means a message received without errors, otherwise, the syndrome value is searched between the columns 57 and 63 of the matrix G, and if this value is founded, a XOR operation between the first 56 bits of the codeword and the 56 bits of the G corresponding line is performed, resulting in the original value of the message (BEZERRA et al., 2010; BEZERRA; AZEVEDO; SILVA, 2011).

Following the TM flow, the data generated by the sensors are processed in the OBDH, which creates a telemetry packet and sends it to the UTMC, then a Telemetry Transfer Frame (TMTF) is generated by the Transfer Layer. The TMTF can be coded with either of Reed-Solomon (RS), Convolutional, or the combination of Reed-Solomon + Convolutional algorithms and results in the CADU that is dispatched by communication interface (BEZERRA et al., 2010; BEZ-ERRA; AZEVEDO; SILVA, 2011).

The implementation of the RS(255, 223) code is based on a Linear Feedback Shift Register (LFSR) that is used for mapping symbols in terms of its basic elements. However, to reduce the area consumption for the add and multiply operations realized by the LFSR, the UTMC applies two pre-defined vectors holding log and anti-log tables for circuit optimization. The convolutional encoding generates two encoded bits for each received bit and performs this action as the last step before to send the data to the UTMC output and follow the mechanism presented in Figure 26.





Source: adapted from (BEZERRA et al., 2010).

3 RELATED WORKS

This chapter presents the related work in the context of communication protocols. In addition, there is a discussion around the system controller, considering its internal resources and the PUS Services usage. A comparison among the related works, highlighting important features of each work, is summarized in a table at the end of the chapter.

Selčan, Kirbiš and Kramberger (2015) present in their work an FPGA-based Communication Stack for nanosatellites using the CCSDS recommendations. The work implements its communication protocols, the TC flow performs the BCH code, and the TM flow implements a concatenated code, which combines Reed-Solomon encoding with the Convolution encoding. Following the CCSDS recommendations, the authors developed a FARM with some modifications targeting the reduction of resource use. However, they do not give a comparison between this modified FARM and the proposed in the CCSDS, in terms of used resources. In order to interface the communications between the systems that apply the Communication Stack, the architecture uses a CAN bus with the CAN 2.0 B protocol. This Communication Stack was designed for the TRISAT mission (TRISAT, 2019).

In Suresh et al. (2015), the authors describe the development of a Command and Data Management System (CDMS) to the nanosatellite IITMSAT. The authors developed the system using a 32-bits ARM microcontroller based on the mbed-RTOS system to schedule the system tasks. As principal function, the controller has as to manage with the generated telemetry data coming from the payload and sensors by I^2C and SPI interfaces. The system also uses an adapted version of the CCSDS protocol, which, notwithstanding the lack of information about the EDAC in the TC flow, the authors describe the use of a convolutional code with a constraint length of 5 in the TM channel coding (MEVADA et al., 2015). This work also implements the following PUS services: Telecommand Verification; Memory Management; Payload Management; Function Management; Onboard Operations Scheduling; Large Data Transfer; and On-Board Storage and Retrieval.

Ivanov et al. (2014) describes the CubETH project, that was developed by the Swiss's Polytechnical School. The system was implemented by using a protocol following the ECSS recommendations (based on earlier recommendations of the CCSDS). In the telecommand flow, the BCH coding scheme was used, as well as, the Reed-Solomon and Convolutional codes in the telemetry flow. The system implementation is made in an MCU (Microcontroller Unit) MSP340. This work also describes a CDMS that acts as the On-Board Computer that has the primary function to get the telemetry information and scheduling the received commands.

The European Space Agency (ESA) is developing a nanosatellite in a CubeSat format which is named OPS-SAT (European Space Agency, 2019). This project proposes the use of CCSDS protocols, the implementations of the communication stack and the coding schemes based on an IP core embedded into a radiation hardened reconfigurable FPGA. Also, the project follows the PUS standard services (EVANS; MERRI, 2014).

In this context, this work proposes the implementation of a communication module into a Rad-Hard FPGA following the CCSDS recommendations. The objective is the usage in nanosatellites missions. This module can be configured to use the BCH scheme into the telecommand flux, as well as the RS, Convolutional, and a combination between RS and Convolutional in the telemetry flux. Thus, the work presents a range of possible EDACs that could be chosen depending on the mission's requirements. Following the CCSDS recommendations, the Communications Operations Procedure (COP) that defines the transmissions frames reliability, has a pertinent part defined as the Frame Acceptance and Reporting Mechanism (FARM), this retransmission mode is used in the implementation.

This work utilizes UART communication interface between the module and others systems (e.g., radio transceiver), it is possible to modify this interface to SSI or SPI, but the second one is limited in the telemetry flow. Also, the communication module applies the Command Pulse Distribution Unit (CPDU), this service is defined in PUS Standard (ECSS-E-ST-70-41C) and enables direct access from the ground to onboard equipment.

Table 3 synthesizes the principal's points of the presented works and aims to facilitate the comparison between them. Fields marked with a dash (-) refer information that was not founded, and fields marked with "N.I." refer to not implemented.

Concerning the presented works, hardware implementation using FPGA, as mentioned in Section 2.2, can potentially achieve higher performance than a microcontroller solution since its remove the overhead of reading/interpreting/executing each instruction that is applied in microcontroller architectures. Besides, the use of a radiation-hardened component increases the reliability of the system once it is subjected to a radiation environment.

The present work uses the CCSDS protocol, including its coding layer. However, we can highlight that this work, besides applying the protocol structure, uses the retransmission mechanism performed by the FARM and implements the structure proposed by the ECSS PUS recommendation in the network layer. The CPDU service is used for direct ground access to subsystems.

Table 3 - Characterization of the systems presented in the related works.

	(Selčan; Kirbiš; Kram- berger, 2015)	(Suresh et al., 2015)	(Ivanov et al., 2014)	(Evans; Merri, 2014)	This Work
Imple- mentation	HW	HW	SW	SW	HW
Target Device	Rad-Hard FPGA	FPGA	MCU	MCU	Rad-Hard FPGA
Protocol	CCSDS	CCSDS	CCSDS	CCSDS (custom)	CCSDS
TM EDAC	RS, Convo- lutional	Convo- lutional	RS, Convo- lutional	-	RS, Convo- lutional
TC EDAC	BCH	-	BCH	-	BCH
Retrans- mission	FARM (custom)	N.I.	N.I.	-	FARM
Com. Interface	CAN	UART	-	SPI	UART
PUS Services	N.I.	Some Services	N.I.	Imple- mented	CPDU

Source: Made by the author, based in: (SELČAN; KIRBIŠ; KRAM-BERGER, 2015),(SURESH et al., 2015),(IVANOV et al., 2014),(EVANS; MERRI, 2014).

4 PROPOSED COMMUNICATION MODULE

4.1 MATERIALS AND METHODS

4.1.1 Hardware Description Language

This work uses VHDL (VHSIC Hardware Description Language) to model digital systems at several levels of abstraction ranging from the algorithm level to the gate level. This language is an integration of: sequential; combinatorial; concurrent; net-list; timing constraints; and waveform generation language (BHASKER, 1999; BEZERRA; LETTNIN, 2014).

Especific CAD (Computer Aided Design) tools use the VHDL language to converter the description that models the system into a stream of bits used to programming the FPGA. The flow to generate the final bit stream to program the FPGA could be roughly divided into five steps: logic synthesis; technology mapping; mapping; placement; and routing (FAROOQ; MARRAKCHI; MEHREZ, 2012).

4.1.2 Finite State Machine Coding

The hardware description uses a set of finite state machine (FSM) to control the circuit. The construction model is based on the XTS User Guide defined by Xilinx (XILINX; INC, 2002), following the standard used in the implementation of the UTMC.

The HDL Coding Technique of FSM implementations is the three process scheme. This model constitutes of three blocks: a "Next State Function" that applies the logic to define the next state based on the actual one and is described by the process 1; a "State Register" which implements a register that stores the current state as a function of the next state and is described by the process 2; and the "Output Function" that is responsible for output assignments and is described by the process 3 (XILINX; INC, 2002; BEZERRA; LETTNIN, 2014). The output function can be registered, and, in this case of a Mealy Machine, the generated outputs are related to inputs.

Figure 27 presents the diagram of this scheme relating the blocks with their process, and a VHDL example description is presented at Appendix A - FSM VHDL Coding Example.



Figure 27 – Finite state machine with three processes diagram.

Source: adapted from (XILINX; INC, 2002).

4.1.3 Development Kit

For the hardware implementation, a development kit board (DevKit), provided by the NanoXplore company, was used.

The DevKit provides means for testing the implementation on BRAVE NG-Medium CLGA625 FPGA, and afford access to an onboard crystal oscillator of 25 MHz and also externals SMA clocks inputs, a 128Mx16 DDR2 memory chip, interfaces such as switches, pushbuttons, LEDs and connectors. Figure 28 presents the development kit.

Figure 28 - NanoXplore Brave development kit.



Source: from author.

4.1.4 Cortex

During the development work, it was possible to have access to Cortex equipment that is a Command Ranging & Telemetry Unit which is a reference for earth observation missions since it has as main features Telemetry data processing and provides CCSDS decoders. This equipment can work with a 70 MHz intermediate frequency, a Doppler estimation and spread spectrum processing for TM and TC.

During the short period of testing, the telemetry flow test was performed with the objective of verifying the stability of the data generated by the implementation. The test setup was composed of the Brave DevKit running the UTMC implementation in a continuous transmission mode. The telemetry output is connected with the Cortex that checks data consistency and performs a de-codification accordingly the used EDAC. Moreover, through a TCP/IP interface, a computer stores the logs from the Cortex. Figure 29 presents the equipment.



Figure 29 – Cortex setup.

Source: (Censin Technology, 2019)

4.2 ON-BOARD DATA HANDLING

The On-Board Data Handling (OBDH) has two main objectives: validation and distribution of received telecommands; and generation and packetization of the generated data. Then, the architecture provides a collection of entities with different functionality being able to identify and decode a received telecommand using CRC algorithms, generate reports by request, and packet the generated data (e.g., I^2C sensors).

The OBDH handles the Network Layer defined by the Section 2.4.3, by using the Space Packet Protocol based on the ECSS PUS definition. The following subsections present the developed architecture and its functionality. To represent the developed OBDH architecture, Figure 30 presents a top-level diagram with the essential connections between the blocks that constitute this system.



Figure 30 – OBDH top level diagram.

Source: from author.

The internal communication between the blocks is based on the use of two embedded memories, one for the telecommand flow and the second for the telemetry flow. The use of these embedded memories allows the exchange of internal information with a smaller number of interconnections and makes the inclusion of new blocks flexible to this architecture since there is no need for significant changes in the existing blocks to have access to the data.

The basic functionality of the blocks are presented as following:

- Packetization Layer: it receives the telecommand data from UTMC and stores the information in the telecommand embedded memory;
- Verification Layer: it verifies the structure of the received telecommand and verifies the data integrity with a CRC algorithm;
- Route Layer: when a telecommand passes through the verification, this layer defines the data destination and distributes the command;
- Configure Layer: it changes under request the value of an internal register (e.g., register that defines the interval time between I²C acquisitions);
- Status Layer: generates by request status reports with information of the internal registers values and reports how many telecommand packets were successfully received and how many telemetry packets were generated;
- Report Layer: it generates, under telecommand request, acknowledgments identifying the success or failure in the reception of a new telecommand; this report informs with an error code in the frame whether a failure occurred;
- I²C Controller: it operates with the I²C interface to acquire data from a set of sensors; this data is stored in the telemetry embedded memory;
- Memory Control: it controls the memory access of the blocks; this layer works with a static priority;
- Transfer Layer: when the amount of generated data is able to fill a complete telemetry packet (accordingly the CCSDS frame structure) or when a request is done, this layer reads the data from the telemetry memory and transfer using the communication interface.

The following subsections present the operation of the blocks that make up the OBDH. In the graphical representations of the implemented state machines, the transitions are made unconditionally. However, the transitions are conditioned to the occurrence of changes in input signals, which are not explicitly shown in the FSMs, in order to simplify the graphic representation.

4.2.1 Packetization Layer

Packetization Layer has the function to store the received telecommand in TC memory. There are two different mechanisms to define the end of a new packet. The first one identifies the information about the packet length, this can be done checking the 5^{th} and 6^{th} byte of the Space Packet Protocol. Then, at each received data, this layer verifies if the packet is finished.

The second mechanism is the utilization of a timeout counter since the information about the packet length can be corrupted. During the receiving process, if the final of a packet is not achieved and the implementation did not receive a new byte during a time window, the process identifies a timeout and this layer informs the Verification Layer that a timeout occur and the packet is discarded. Figure 31 presents a simplified FSM of this procedure.



Figure 31 – Packetization layer finite state machine.

Source: from author.
4.2.2 Verification Layer

The Verification Layer starts its procedure when the Package process is complete. The first step is to check if a timeout error has occurred in the Packetization layer, and, when true, a report is generated with a timeout error, and the packet is discarded. If no timeout error occurs, the systems begin to obtain the data and verify if the fields that made up the structure of a space packet are consistent and starts the CRC algorithm.

When all data are verified, the FSM can follow different flows. If the CRC and data are consistent, this layer will report a ready signal to the Route Layer, and if there is a request, a report will be generated. If the CRC is incorrect, a report is generated, and the package is discarded. If a fielding error has occurred, the report is generated if requested, and the packet is discarded. Figure 32 presents the FSM of this layer.



Figure 32 – Verification layer finite state machine.

4.2.3 Route Layer

Since the Verification layer sends a ready signal, the Routing layer gets the telecommand data and identifies the destination. Next, the process routes a ready signal to its destination and wait for an acknowledgment. Figure 33 shows the FSM applied in this process.

Figure 33 – Route layer finite state machine.



Source: from author.

4.2.4 Configure Layer

Configuration layer is capable of reconfiguring the value of an internal register, for example, the register that defines the acquisition period of the I^2C Controller, the procedure is simple, and this entity is able of modifying the register when requested. Figure 34 presents an FSM to show this procedure.



Figure 34 – Configuration layer finite state machine.

Source: from author.

4.2.5 Status Layer

The status layer is able of creating a packet with information about the internal configurable registers and informing how many successfully packets are received and sent. The report follows the structure defined in the ECSS PUS with an error control field that is filled by a 16 bits CRC. Figure 35 presents the FSM that defines the process.



Figure 35 – Status layer finite state machine.

4.2.6 Report Layer

The Report Layer creates an ACK or NACK report according to the error code received from the Verification layer. As the previous blocks, the Report follows the ECSS PUS structure and maintains a different counter for ACK and NACK packets. The Report layer process is presented in Figure 36.



Figure 36 – Report layer finite state machine.

4.2.7 Transfer Layer

The transfer layer regularly checks the amount of data generated by the sensors and requests from the Report and Status layers. Then, when the sensors data achieve a size to fill a full telemetry packet, or a request event occurs, this layer read the data from the TM memory and transfers it through the communication interface. The FSM applied in Transfer layer is showed in Figure 37.

Figure 37 – Transfer layer finite state machine.



Source: from author.

4.2.8 Memory Control

Memory Controller acts as an interface between the blocks that access the embedded memory. When the process receives request signals the access is given to the block with highest priority. As the procedures to receive, verify, route and act from a telecommand are made in stages, the request signals are needed at different times, and there is no need of a dynamic priority algorithm. However, in the TM flow, the I²C Controller accesses the memory regularly and to allow Report and Status blocks to access memory, which requires sporadic access. This layer has the lowest priority in the system.

In addition to the priority access mechanism, this layer applies the write and read commands according to the address (read procedure) and data (read and write procedure) for all blocks connected to it. Figure 38 presents a simplified state machine that explicates the process sequence.



Figure 38 – Memory control finite state machine.

4.2.9 I²C Controller

The I²C Controller interacts with a communication interface and controls the flux to receive data from all the sensors connected in the bus. The layer stores the acquired data in the embedded memory according to an interval time defined by an internal register (the same that can be modified by the Configuration layer).

Figure 39 presents the flow of this process using a FSM. This layer can be easily modified to achieve different procedures in the reading protocol of the devices. The actual architecture works with five temperature sensors and a total-elapsed-time recorder (TET-R).



Figure $39 - I^2C$ controller finite state machine.

Besides the use of the Cortex equipment, a test setup was used to check the correct flow of telecommands and telemetry using the FPGA implementation. The proposed setup uses a USB-RS232 converter, the NanoXplore development kit and python scripts are responsible for the data flow control between a host computer and the DUT (Device Under Test).

In this test context, valid and invalids CLTUs are generated by a python script according to the test needs. Then the data are sent by the USB Serial converter using a UART interface with a transmission speed of 115200 bit/s. The received CLTU is transmitted to the UTMC and from there different results can be achieved. Since the telecommand can have different destinations, when a data is expected as an answer, the Python script is able to get the data from the UTMC and plot according to the CCSDS packets structures. Figure 40 presents the proposed setup.



Figure 40 – Test setup

Source: from author.

4.4 PAYLOAD-X

In the context of this research, a payload was developed for the FloripaSat mission, with the purpose of in-orbit validation of the communication module and a hardware configuration update mechanism. This project is named Payload-X and Figure 41 shows the project interconnection diagram.



Figure 41 – Payload-X interconnection diagram.

Source: from (LUZA et al., 2018; RIGO et al., 2019).

The printed circuit board (PCB) where the architecture is implemented was designed by Cézar A. Rigo as a result of his master project (RIGO, 2019), the design follows the European Space Agency (ESA) space product standards. It has a layered structure that mitigates the effects of radiation and electromagnetic interference on the components signals. All signal layers lie between power and ground planes, avoiding tracks on the outer layers. Moreover, all components were selected to tolerate wide temperature variation, and some of then can tolerate radiation – as the microcontroller (MSP430FR6989) with ferromagnetic program memory and the rad-hard FPGA (NX1H35S-BG625PR) (LUZA et al., 2018; RIGO et al., 2019).

A central feature of the architecture is its capability to change the hardware configuration of the FPGA through a remote up-link of its bit stream. The proposed mechanism is called HUMAN (Housekeeper and Update MANager) and was proposed in the master project from Elder D. Tramontin (TRAMONTIN, 2018). The MCU is responsible for updating the configuration bistream stored in a non-volatile flash memory. An alternative bitstream is also stored in the memory, for applying a fail-safe technique. There are three stored bitstreams copies, and a voting scheme is used to ensure data integrity, since the flash memory is susceptible to SEE. The MCU module is also responsible for the housekeeping and for the update management of the radiation hardened FPGA. The architecture also includes five temperature sensors, a TET-R and three SRAM devices. The data collected by these devices will be used for radiation monitoring (LUZA et al., 2018; RIGO et al., 2019).

In the present work, it has been developed the application stored in the FPGA, including the proposed OBDH, and the the UTMC porting to the BRAVE FGPA. The communication module handles TC and TM data, and it is an interface between the radio transceiver and the OBDH. The OBDH is based on the CCSDS and ECSS recommendations and performs the validation of the received telecommand, and the packeting of the telemetry that is acquired by the available sensors.

Figure 42 presents these works contribution overview.



Figure 42 – Contribution for the Payload-X project.

Source: from author.

Figure 43 presents the model of the board.



Figure 43 – Payload-X board.

Source: from (LUZA et al., 2018; RIGO et al., 2019).

4.5 FLORIPASAT-I INTEGRATION

The payload architecture presented in the previous section is under integration in the FloripaSat mission of the Federal University of Santa Catarina. The FloripaSat-I satellite consists of a 1U Cube-Sat with five functional modules: EPS; OBDH; TT&C; battery and interface board, besides the payload modules where the Payload-X is integrated (SLONGO et al., 2016).

Payload-X integration with FloripaSat-I is based in two operation modes. The first mode is called Nominal and it uses the I^2C interface available to communicate with the FloripaSat-I OBDH. It is treated as a safe operating mode. The second mode, called Advanced, uses the SPI interface and GPIO to control the CubeSat radio during a window of 4 hours or while the energy level is at higher status, and it is used as a risky operation mode once it takes control of CubeSat's main communication channel. The following subsections will define both operation modes and the data flow related to case scenarios.

4.5.1 Nominal Operation Mode

The Nominal operation mode was designed to use the interface communication already available in FloripaSat-I's architecture since it has an I²C channel to communicate with the payload. The UTMC does not have an I²C interface, so the microcontroller is used to convert this I²C to a UART communication interface. The microcontroller also uses its FRAM memory to store telemetry received from the FPGA implementation, then, when the FloripaSat-I OBDH requires data from the Payload-X, the microcontroller sends the last received telemetry to be stored by the FloripaSat-I. Figure 44 shows the blocks diagram of Nominal mode.

4.5.2 Advanced Operation Mode

The Advanced mode was designed to make use of the CCSDS and ECSS protocols directly, bypassing the NGHam protocol used by the FloripaSat-I communication module. A telecommand starts this mode, and when the FloripaSat-I recognizes this telecommand, it allows the Payload-X to assume full control of the radio transceiver for a 4 hour time window. Figure 45 presents the diagram of Advanced mode.



Figure 44 – Nominal operation mode top level diagram.

Source: from author.

Figure 45 – Advanced operation mode top level diagram.



Source: from author.

4.5.3 Data Flow

For each case, there is a diagram showing the respective data flow, and there is a step-by-step description of the performed activities. The acronym FS will be used to referre to FloripaSat-I, and the acronym PX will be used for Payload-X.

4.5.3.1 CCSDS Telecommand

For the validation of the CCSDS communication, it is necessary to send some data or telecommand over CCSDS protocol. Until reaching the UTMC (CCSDS decoder), the packet should pass through other devices. The Figure 46 shows the data flow of the CCSDS telecommand case. The activities performed in each step are described in Table 4.

4.5.3.2 CCSDS Telemetry

As well as the CCSDS Telemetry, it is necessary to receive some data over CCSDS protocol to validate the CCSDS communication. Some events could generate telemetry: a telecommand or a measurement from Payload OBDH. The Figure 47 shows the data flow of the CCSDS telemetry case. The activities performed in each step are described in the Table 5.



Figure 46 – Use case CCSDS telecommand diagram.

Source: from author.

Step	Description	Source	Destination				
1	Ground station sends data @	Ground-	FS-TT&C				
	UFH to FloripaSat-I	station					
2	FS-OBDH read data from ra-	FS-TT&C	FS-OBDH				
	dio						
	FS-OBDH decodes NGHam	FS-OBDH	FS-OBDH				
	packet						
	FS-OBDH interprets the TC	FS-OBDH	FS-OBDH				
3	FS-OBDH request to send	FS-OBDH	PX-HUMAN				
	TC data to PX						
	FS-HUMAN stores data on	PX-HUMAN	PX-HUMAN				
	the FRAM memory						
4	PX-HUMAN sends TC to	PX-HUMAN	PX-UTMC				
	PX-UTMC						
	PX-UTMC decodes CLTU	PX-UTMC	PX-UTMC				
	frame						
	PX-UTMC decodes TC	PX-UTMC	PX-OBDH				
	frame						
	PX-OBDH decodes TC	PX-OBDH	PX-OBDH				
	Packet						
	PX-OBDH extracts usefull	PX-OBDH	PX-OBDH				
	data						

Table 4 – CCSDS Telecommand step-by-step.





Source: from author.

Step	Description	Source	Destination
1	PX-OBDH generated a TM	PX-OBDH	PX-UTMC
	packet		
	PX-UTMC sends a CCSDS	PX-UTMC	PX-HUMAN
	TM packet to PX-HUMAN		
	PX-HUMAN stores the data	PX-HUMAN	PX-HUMAN
	in FRAM		
2	When received a request,	PX-HUMAN	FS-OBDH
	PX-HUMAN sends the data		
	to FS-OBDH		
3	FS-OBDH decodes each	FS-OBDH	FS-TT&C
	packet in NGHam protocol		
	and send to FS-TT&C		
4	FS-TT&C send data @ UHF	FS-TT&C	Ground-
	to Ground station		station

Table 5 – CCSDS Telemetry step-by-step.

Source: from author.

4.5.3.3 Bitstream Upload

When a new bitstream needs to be uploaded to Payload X, it is split into many segments and sent through the radio. From the ground station until data reaching the target in the Payload X board, some control bytes are added to the segments. Figure 48 shows the data flow of a bitstream segment upload.





Source: from author.

Sten	Description	Source	Destination
1	Ground station sends data @	Ground-	FS-TT&C
T	UHE to Elorino Set I	station	10-1100
0	ES ODDU and data from an		EC ODDI
Ζ	FS-OBDH read data from ra-	FS-11&C	FS-OBDH
	d10		
	FS-OBDH decodes NGHam	FS-OBDH	FS-OBDH
	packet		
	FS-OBDH interprets the	FS-OBDH	FS-OBDH
	telecommand		
3	FS-OBDH request to send	FS-OBDH	PX-HUMAN
	bitstream data to Payload-X		
	PX-HUMAN stores data on	PX-HUMAN	PX-HUMAN
	the FRAM memory		
	PX-HUMAN check message	PX_HUMAN	PX_HIIMAN
	integrity using CPC	1 74-110 1017110	1 7-110 1017110
4	DY HUMAN	DV IIIMAN	
4	PX-HUMAN stores data at	PX-HUMAN	Flash-
	3 positions on the flash me-		memory
	mory		

Table 6 – Bitstream upload step-by-step.

4.5.3.4 Bitstream Status Request

Since the bitstream upload has no acknowledgment, to know if the packets reached correctly the target, it is necessary to send a "Status Request" telecommand. The Figure 49 shows the data flow of a bitstream status request.

Figure 49 – Use case bitstream status request diagram.



Source: from author.

Table 7 – Bitstream status request step-by-step.

Step	Description	Source	Destination		
1	Ground station sends data @	Ground-	FS-TT&C		
	UHF to FloripaSat-I	station			
2	FS-OBDH read data from ra-	FS-TT&C	FS-OBDH		
	dio				
	FS-OBDH decodes NGHam	FS-OBDH	FS-OBDH		
	packet				
	FS-OBDH interprets the	FS-OBDH	FS-OBDH		
	telecommand				
3	FS-OBDH request bit-	FS-OBDH	PX-HUMAN		
	stream status data from				
	PX-HUMAN				

4.5.3.5 Bitstream Status Reply

Once the status request arrives at the HUMAN, it is necessary to answer. Since bitstream segment status is too large to fit into 1 data package, it needs to be split into many packets and, then, send over the radio. Figure 50 shows the data flow of a bitstream status reply.



Figure 50 – Use case bitstream status reply diagram.

Source: from author.

Table 8 – Bitstream status reply step-by-step.

Step	Description	Source	Destination		
1	Ground station sends data @	Ground-	FS-TT&C		
	UHF to FloripaSat-I	station			
2	FS-OBDH read data from ra-	FS-TT&C	FS-OBDH		
	dio				
	FS-OBDH decodes NGHam	FS-OBDH	FS-OBDH		
	packet				
	FS-OBDH interprets the	FS-OBDH	FS-OBDH		
	telecommand				
3	FS-OBDH requests from	FS-OBDH	PX-HUMAN		
	Payload-X to swap version				
	PX-HUMAN change the flag	FS-HUMAN	FS-HUMAN		
	that indicates the current				
	version in use				

4.5.3.6 Bitstream Swap Version

There are always two bit stream versions on-board the Payload-X: the currently used version and the alternative version (fallback). To swap from the latter to the former, it is necessary to send a "Bitstream Swap Version" telecommand. Figure 51 shows the data flow of a bitstream swap version.





Source: from author.

Table 9 – Bitstream swap version step-by-step.

Step	Description	Source	Destination		
1	PX-HUMAN reads the frame	PX-HUMAN	Flash-		
	status from flash memory		memory		
2	PX-HUMAN sends status	PX-HUMAN	FS-OBDH		
	data to FS-OBDH				
	FS-OBDH encodes NGHam	FS-OBDH	FS-OBDH		
	packet				
3	FS-OBDH writes data on ra-	FS-OBDH	FS-TT&C		
	dio				
4	FloripaSat-I sends data @	FS-TT&C	Ground-		
	UHF to Ground station		station		

5 RESULTS

5.1 UTMC PORTING TO BRAVE FPGA

The implementation of the UTMC in the Brave FPGA required modifications over the original VHDL description since the NanoXplore synthesis tool infers logical blocks according to the resources available in the FPGA and the result of the RTL is different from those presented by tools such as Quartus, Vivado, and Libero (which was the tool used in the original UTMC design).

The two encountered problems were the inference of latches in state machine implementations and low operating frequency.

The solution for the first problem is directly connected with the way the synthesis tool works. The described process has to cover all the possibilities, then, the "case others" process statement needed a function. Using this need, if the process achieves the "others =>" statement it means that an error occurred in the transition from the current state to the new state, thereby this implementation uses this process to generate an error signal to force to reset in the FPGA.

Also, some processes need to be modified to avoid latch inference. An example of this modification is presented next.

```
-- Process with latch
process : process (input)
   begin
        if (input = '1') then
           state <= s1;</pre>
        end if:
    end process process;
-- Process without latch
process : process (input)
    begin
        if (input = '1') then
           state <= s1:
        else
           state <= s2;</pre>
       end if:
    end process process;
```

The approach related to the second issue is associated with the way the codification process has been implemented. The BCH implementation used a "for" statement to do the cyclic interactions, then, it generated a huge combinational logic between registers and created a long critical path. Figure 52 presents the problematic. The "Data delay" must be lower than the "Clock delay", then, with a large combinational logic, the maximum operation frequency is 8.5 MHz.

Figure 52 – Data and clock delay between a source and destination register.



Source: adapted from (NANOXPLORE, 2018b)

Then, the approach used to avoid this issue was to add pipeline, thus the "for" procedure was split in several stages reducing the combinational logic between the registers. Although, with this approach, the number of cycles were increased to finalize the operation, it was possible to achieve a four times higher operating frequency. This modification is necessary since the resources made available by the BRAVE FPGA are different from those of the original application, since in the original application the operating frequency was higher than required.

5.2 USE CASES

This section shows the results of the use cases of the system; each subsection presents its telecommand structure and also the process using the python script. Moreover, a bit flip is randomly inserted in all the constructed CLTUs, as a fault injection mechanism. The fault injection system is not detailed in the manuscript since it is out of the scope of this work.

5.2.1 OBDH Register Configuration Flow

The operation consists of sending a specific command that defines the address of the registers and the configuration values. Constructed from the data presented in the Figure 53, the command generates the CLTU, which includes all the CCSDS layers used in this implementation. The generated CLTU is sent from a python script that operates as serial USB converter, following the setup test presented in Section 4.3.

This command should generate a system response since a return is requested using the "Ack Flags" field. It is interesting to highlight that the "Ack Flags" field of the "TC Packet Secondary Header" is set as "b1000" meaning that a verification report has to be generated.

Figure 53 –	OBDH	register	configuration	telecommand	structure.
0		0	0		

•			тс	C Transi	fer Fran	ne Pri	mary Hea	der-				
Version Number 2 bits	Bypas Flag 1 bit	s	Control Command Flag 1 bit	Rese Sp 2 t	served Spacecraft Channel Identifier bits 10 bits 6 bits		Frame Length 10 bits		Frame Sequence Number 8 bits			
b00	b0		0b	b	b00 b111111111			b	000001	b000001	0110	x64
•				Pack	et Prima	ary H	eader					>
]	Packet Identific	ation			Pack	et Se	quence Co	ontrol	1	Packet
Packet Version Number 3 bits	Packe Type 1 bit	t	Secondary Header Flag 1 bit	Ap P Id	plication Process lentifier 11 bits	n	Sequence Flags 2 bits		Packet Sequence Count or Packet Name 14 bits		Data Length 16 bits	
b000	b1		b1 b000000			00	b11	b11 b00000		00000000		x0008
•			TC Pac	ket Sec	condary	Head	ler		·····>			
			Ν	Aessage	e Type I	D				1		
TC PUS Version Number 4 bits	TC PUS Version Number 4 bits Ack Flags 4 bits		Service Typ 8 bits	Message Subtype ID 8 bits		Source ID 16 bits						
b0010	b100)	x30			x0	1	:	x0000			
•	- User Da	ta Fi	eld	>								
	Source	e Dat	a									
Register A 8 bits	ddress		Register Value 8 bits	:	CRC 16 bits		FEC 16 bit		s			
x01			xF0		:	x556	6		xEB7	4		

A previous simulation is done to verify the behavior of the system, the main signals related with this procedure are presented in Figure 54.

Figure 54 – Results captured from the VHDL simulation of the OBDH register configuration flow (ModelSim).



Source: from author.

The Configure Layer starts the configuration process when the signal "route_ready_i" changes from '0' to '1'. The pointer address presented in the vector signal "route_data_i" indicates the data location in the embedded memory, and from these data the process identify the destination "config_register_sv" and the new value "config_value_sv". Then, the destination receives the new value and returns an acknowledged to the layer showing a successful configuration.

As a result of this simulation, the Report Layer creates a report packet following the space packet structure. A positive acknowledged showed by the signal " $flag_ack_s$ " (that is is a high state) that embeds the "error_code_sv" information (which in case of a positive acknowledged, is defined as "000h") and a 16-bit CRC "crc_out_sv" is generated.

The signal "*report_ready_i*" indicates that the Report Layer procedure is finished and the Transfer Layer starts to send the data through the UART interface for the UTMC.

Figures 55, 56 and 57 present the telecommand data and the received telemetry by using the test setup. Unlock and Set commands are needed to handle the FARM implementation. In Figure 56 the "Service Type ID" identifies that it is a verification report, and the "Message Type ID" indicates that is a ACK message. The "User Data" presents the Space Packet primary header of the source telecommand packet.

Figure 55 – OBDH register configuration flow: telecommand to be sent.

lucas@cubos-licenses:~/Projects/payload-x/scripts\$ python3 test_procedure.py Payload-X Test Procedure => Choose the use case: 1 - Destination: OBDH Config | Expected: Ack 2 - Destination: OBDH Status | Expected: Status Report 3 - Destination: UTMC CPDU | Expected: Ack 4 - Destination: UTMC CPDU | Expected: NAck (code error x08) 5 - Destination: OBDH Expected: NAck (code error x080) Commando: 1 => Telecommand (TT&C -> Payload-X): Unlock Command Generated CLTU: Start Sequence (2 bytes): xEB90 BCH Code Block (8 bytes): x33FF04070000FA52 BCH Code Block (8 bytes): xB2000000000008A Stop Sequence (8 bytes): xC5C5C5C5C5C5C579 Set Command Generated CLTU: Start Sequence (2 bytes): xEB90 BCH Code Block (8 bytes): x33FF0409008200EE BCH Code Block (8 bytes): x64CD23000000014 Stop Sequence (8 bytes): xC5C5C5C5C5C5C579 __Configure Command__ Generated CLTU: Start Sequence (2 bytes): xEB90 BCH Code Block (8 bytes): x03FF041664C018BE BCH Code Block (8 bytes): x00C0000082830A0 BCH Code Block (8 bytes): x01000001F05566BA BCH Code Block (8 bytes): xEB740000000002A Stop_Sequence (8 bytes): xC5C5C5C5C5C5C579 Source: from author.

Figure 56 – OBDH register configuration flow: expected telemetry (first part).

```
=> Telemetry (Payload-X -> TT&C):
ASM: x1acffc1d
Transfer Frame Primary Header: x3ff300001800
  .Version Number (2 bits): b00
  .Virtual Channel ID (3 bits): b001
  .Operational Control Field Flag (1 bit): b1
  .Master Channel Frame Counter (1 byte): x00
  .Virtual Channel Frame Counter (1 byte): x00
  .Transfer Frame Data Field Status (2 bytes): x1800
Space Packet Primary Header: x0800c0000010
  .Packet Version Number (3 bits): b000
  .Packet Type (1 bit): b0
  .Application Process ID (11 bits): b0000000000
  .Sequence Flag (2 bits): b11
  .Sequence Count (14 bits): xc000
  .Packet Lenght (2 bytes): x0010
Space Packet Secondary Header: x200101000f00000000000
  .TM PUS Version Number (4 bits): b0010
  .Spacecraft Time (4 bits): b0000
  .Service Type ID (1 byte): x01
  .Message Type ID (1 byte): x01
  .Message Type Counter (2 byte): x000f
  .Destination ID (2 bytes): x0000
  .Time (1 byte): x00
  .Spare (3 byte): x000000
User Data : x1800c000
CRC (2 bytes): xeed6
```

Figure 57 – OBDH register configuration flow: expected telemetry (second part).

```
Idle Packet: x07ffc000043155..55
  .Packet Version Number (3 bits): b000
  .Packet Type (1 bit): b0
  .Secondary Header Flag (1 bit): b0
  .Application Process ID (11 bits): b1111111111
  .Sequence Flag (2 bits): b11
  .Sequence Count (14 bits): xc000
  .Packet Lenght (2 bytes): x0431
  .Idle Data: x55..55
CLCW: x01040465
  .Type (1 bit): b0
  .Version Number (2 bits): b00
  .Status (3 bits): b000
  .COP Use (2 bits): b01
  .Virtual Channel ID (6 bits): b000001
  .Reserved (2 bits): b00
  .Flags (5 bits): b00000
  .FARM-B Counter (2 bits): b10
  .Reserved (1 bit): b0
  .Informed Value (1 byte): b65
FECW (2 bytes): x177e
       __Test Procedure Finished_____
     Source: from author.
```

5.2.2 OBDH Status Request Flow

The OBDH system has the capability to transmit a status telemetry frame under request. Following the same test procedure, the interface communication sends a request telecommand to the communication module, and the python script checks the received telemetry consistency.

Figure 58 presents the structure and data of the request telecommand, it is interesting to highlight that the "Ack Flags" field of the "TC Packet Secondary Header" is set to "b00000", mean that a verification report should not to be generated since the answer is the status telemetry.

Figure 58 – OBDH status request telecommand structure.

Version Number 2 bits	Bypass Flag 1 bit	Control Command Flag 1 bit	Reser Spa 2 bi	rved Spacecraft are Identifier its 10 bits		SpacecraftVirtual ChannelFran LengIdentifierIdentifierIdentifier10 bits6 bits10 b		cecraft ntifier bits Virtual Channel Identifier 6 bits		e h s	Fram Sequer Numb 8 bit
b00	b0	0b	b0	00	b11	11111111	b	000001	b0000010)100 x64	
			Packe	et Prima	ry H	eader					>
		Packet Identific	ation			Pack	et Se	quence Co	ontrol	T	Dacket
Packet Version Number 3 bits	Packet Type 1 bit	Secondary Header Flag 1 bit	App Pr Ide	olication rocess entifier 1 bits	1	Sequen Flags 2 bits	ice s	Packet S Cour Packet 14	bequence Data nt or Length Name 16 bits		
b000	b1	b1	ь0000000000			b11		ь000000	00000000	,	x0006
		TC Pac	ket Seco	ondary l	Head	ler		·····>			
		Ν	Aessage	Type II)						
TC PUS Version Number 4 bits	Ack Flags 4 bits	Service Typ 8 bits	e ID	Messag	essage Subtype ID 8 bits		ubtype ID Sourc 16 b				
b0010	b0000	x31			x01 x			x0000			

CRC	FEC
16 bits	16 bits
xD5CA	xD12C

A simulation of this flow is presented in Figure 59.

Figure 59 – Results captured from the VHDL simulation of the OBDH status request flow (ModelSim).



Source: from author.

When the Status Layer receives a request signal route_ready_i from the Route Layer, a status packet is generated within the data get from "tc_counter_sv", "tm_counter_sv", "config_register_1_sv", and "config_register_2_sv" and the 16-bit CRC generated. When the status process is finished, the "status_data_i" is set high and the Transfer Layer starts to sent the packet to the UTMC using the UART interface. Figures 60, 61 and 62 present the telecommand and the telemetry using the test setup.

Figure 60 – OBDH request status flow: telecommand to be sent.

lucas@cubos-licenses:~/Projects/payload-x/scripts\$ python3 test_procedure.py _Payload-X Test Procedure_____ => Choose the use case: 1 - Destination: OBDH Config | Expected: Ack 2 - Destination: OBDH Status | Expected: Status Report 3 - Destination: UTMC CPDU | Expected: Ack 4 - Destination: UTMC CPDU | Expected: NAck (code error x08) Expected: NAck (code error x080) 5 - Destination: OBDH Commando: 2 => Telecommand (TT&C -> Payload-X): Unlock Command Generated CLTU: Start Sequence (2 bytes): xEB90 BCH Code Block (8 bytes): x33FF04070000FA52 BCH Code Block (8 bytes): xB200000000008A Stop Sequence (8 bytes): xC5C5C5C5C5C5C579 ___Set Command___ Generated CLTU: Start Sequence (2 bytes): xEB90 BCH Code Block (8 bytes): x33FF0409008200EE BCH Code Block (8 bytes): x64CD23000000014 Stop Sequence (8 bytes): xC5C5C5C5C5C5C579 ___Status Command___ Generated CLTU: Start Sequence (2 bytes): xEB90 BCH Code Block (8 bytes): x03FF041464C0182A BCH Code Block (8 bytes): x00C0000062031DA BCH Code Block (8 bytes): x01000074D5CAD12C Stop Sequence (8 bytes): xC5C5C5C5C5C5C579 Source: from author.

In Figure 61, the value '31' in the "Service Type ID" identifies that a message is from the Status unit and the "Message Type ID" indicates that the message is a Status Report. The "User Data" field is composed of four bytes, the first byte gives the telemetry counter state, the second byte gives the telecommand counter state, the third byte is the value of the register 1 and the last byte is the value of the internal register 2.

Figure 61 – OBDH request status flow: expected telemetry (first part).

```
=> Telemetry (Payload-X -> TT&C):
ASM: x1acffc1d
Transfer Frame Primary Header: x3ff30e0e1800
  .Version Number (2 bits): b00
  .Virtual Channel ID (3 bits): b001
  .Operational Control Field Flag (1 bit): b1
  .Master Channel Frame Counter (1 byte): x0e
  .Virtual Channel Frame Counter (1 byte): x0e
  .Transfer Frame Data Field Status (2 bytes): x1800
Space Packet Primary Header: x0800c00e000e
  .Packet Version Number (3 bits): b000
  .Packet Type (1 bit): b0
  .Application Process ID (11 bits): b0000000000
  .Sequence Flag (2 bits): b11
  .Sequence Count (14 bits): xc00e
  .Packet Lenght (2 bytes): x000e
Space Packet Secondary Header: x203102000c0000000
  .TM PUS Version Number (4 bits): b0010
  .Spacecraft Time (4 bits): b0000
  .Service Type ID (1 byte): x31
  .Message Type ID (1 byte): x02
  .Message Type Counter (2 byte): x000c
  .Destination ID (2 bytes): x0000
  .Time (1 byte): x00
  .Spare (1 byte): x00
User Data : x00010AFF
```

Figure 62 – OBDH request status flow: expected telemetry (second part).

```
Idle Packet: x07ffc000043355..55
  .Packet Version Number (3 bits): b000
  .Packet Type (1 bit): b0
  .Secondary Header Flag (1 bit): b0
  .Application Process ID (11 bits): b1111111111
  .Sequence Flag (2 bits): b11
  .Sequence Count (14 bits): xc000
  .Packet Lenght (2 bytes): x0433
  .Idle Data: x55..55
CLCW: x01040465
  .Type (1 bit): b0
  .Version Number (2 bits): b00
  .Status (3 bits): b000
  .COP Use (2 bits): b01
  .Virtual Channel ID (6 bits): b000001
  .Reserved (2 bits): b00
  .Flags (5 bits): b00000
  .FARM-B Counter (2 bits): b10
  .Reserved (1 bit): b0
  .Informed Value (1 byte): b65
FECW (2 bytes): x7d4d
     Test Procedure Finished
```

5.2.3 CPDU Command Flow

The Command Pulse Distribution Unit is a simple onboard unit that is capable of providing direct access from the ground to equipment. The flow starts with a telecommand that embeds information about the pulse duration and the target output. A direct telecommand to the CPDU is created within a CLTU; then this data is transferred by the communication interface to the FPGA. The FPGA receives this command and generates an ACK (acknowledged) or NACK (not acknowledged) within a CADU, that is dispatched by the TM flow.

This flow was simulated with ModelSim, and the resulting waveform is presented in Figure 63. For the TC flow, the "start_seq_ok_s" signal is active when the start sequence of the CLTU is identified and the signal "stop_seq_ok_s" is activated by a stop sequence. In the specific example, the CLTU is composed of several blocks that are coded following the BCH(63,56) algorithm, then the signals "status_s" and "aff_s" are defined according to the situations presents in Table 10. Also, the signal "sindrome_s" is responsible for the BCH algebraic decoding method known as syndrome decoding (MOON, 2005). For example, in the flow presented in Figure 63 two errors were found and corrected.

In addition, it is possible to identify the pulse of 13 ms in the signal "pulses_o(0)". The "ready_i" stands for the CPDU acknowledge value, "done_o" is set to '0' when the layer is busy, " $cpdu_size_i$ " represents 13×2^n ms pulse duration and " $cpdu_addr_i$ " is the physical address of the output pin.

Status	Aff	Result
0	Х	An error has been found and can not be corrected
1	0	No error was found
1	1	An error was found and corrected
õ		

Table 10 - BCH cases according the *status* and *aff* signals.

Source: adapted from (BEZERRA et al., 2010).

This test is performed by using the proposed test setup, where a CPDU command is sent through the serial communication interface, and the expected data is checked. The structure of the telecommand and the received ack/nack are respectively presented in the Figure 64, 65, 66 and 67. Figure 63 – Results captured from the VHDL simulation of the CPDU procedure (ModelSim).



Source: from author.

Figure 64 – CPDU telecommand structure.

∢			·TC	Transfer Fran	ne Pri	mary Hea	der-)
Version Number 2 bits	Bypas Flag 1 bit	s Contr Comma Flag 1 bit	ol and	Reserved Spare 2 bits	eserved Spaced Spare Identi 2 bits 10 b		C Id	Virtual Fram hannel Lengt entifier 10 bit 6 bits		e h s	Frame Sequen Numbe 8 bits	er
b00	b0	0b		b00 b11		11111111	b	000000	b000001	0001	xC8	
∢				Packet Prim	ary H	eader					>	
	Packet Identification					Packet Sequence Control Packet					Packet	
Packet Version Number 3 bits	Packe Type 1 bit	Secondary Application Packet Header Process Type Flag Identifier 1 bit 1 bit 11 bits					Packet Sequence Sequence Count or Flags Packet Name 2 bits 14 bits			I	Data Length 6 bits	
b000	b1	b0		ь000000000	01	b11		ь000000	00000000	,	x0003	
∢		User Data Fie	ld		>							
		Source Data										
CPDU Output 12 bit	t Number s	Reserved 1 bit		Register Value 3 bits		1	CRC 6 bit	ts	1	FEC 6 bit	s	
b0000000	00000	b0		ь000		x1D0F			x9C07			
Figure 65 – CPDU command flow: telecommand to be sent.

lucas@cubos-licenses:~/Projects/payload-x/scripts\$ python3 test_procedure.py Pavload-X Test Procedure => Choose the use case: 1 - Destination: OBDH Config | Expected: Ack 2 - Destination: OBDH Status | Expected: Status Report 3 - Destination: UTMC CPDU | Expected: Ack 4 - Destination: UTMC CPDU Expected: NAck (code error x08) 5 - Destination: OBDH | Expected: NAck (code error x080) Commando: 3 => Telecommand (TT&C -> Payload-X): Unlock Command Generated CLTU: Start Sequence (2 bytes): xEB90 BCH Code Block (8 bytes): x33FF0007000030AA BCH Code Block (8 bytes): x430000000000078 Stop Sequence (8 bytes): xC5C5C5C5C5C5C579 Set Command Generated CLTU: Start Sequence (2 bytes): xEB90 BCH Code Block (8 bytes): x33FF000900820052 BCH Code Block (8 bytes): xC8BFE4000000004A Stop Sequence (8 bytes): xC5C5C5C5C5C5C579 ___CPDU Command___ Generated CLTU: Start Sequence (2 bytes): xEB90 BCH Code Block (8 bytes): x03FF0011C8C0108C BCH Code Block (8 bytes): x01C000000300002E BCH Code Block (8 bytes): x1D0F9C070000006C Stop Sequence (8 bytes): xC5C5C5C5C5C5C579 Source: from author.

In Figure 66, the value "01" in the "Service Type ID" identifies that a confirmation message and the value '01' in the "Message Type ID" indicates that the message is ACK report.

Figure 66 – CPDU command flow: expected telemetry (first part).

```
=> Telemetry (Payload-X -> TT&C):
ASM: x1acffc1d
Transfer Frame Primary Header: x3ff300001800
  .Version Number (2 bits): b00
  .Virtual Channel ID (3 bits): b001
  .Operational Control Field Flag (1 bit): b1
  .Master Channel Frame Counter (1 byte): x00
  .Virtual Channel Frame Counter (1 byte): x00
  .Transfer Frame Data Field Status (2 bytes): x1800
Space Packet Primary Header: x0801c000000f
  .Packet Version Number (3 bits): b000
  .Packet Type (1 bit): b0
  .Application Process ID (11 bits): b0000000001
  .Sequence Flag (2 bits): b11
  .Sequence Count (14 bits): xc000
  .Packet Lenght (2 bytes): x000f
Space Packet Secondary Header: x200101000000000
  .TM PUS Version Number (4 bits): b0010
  .Spacecraft Time (4 bits): b0000
  .Service Type ID (1 byte): x01
  .Message Type ID (1 byte): x01
  .Message Type Counter (2 byte): x0000
  .Destination ID (2 bytes): x0000
  .Time (1 byte): x00
User Data : x1001c000
CRC (2 bytes): x0242
     Source: from author.
```

Figure 67 – CPDU command flow: expected telemetry (second part).

```
Idle Packet: x07ffc000043455..55
  .Packet Version Number (3 bits): b000
  .Packet Type (1 bit): b0
  .Secondary Header Flag (1 bit): b0
  .Application Process ID (11 bits): b1111111111
  .Sequence Flag (2 bits): b11
  .Sequence Count (14 bits): x0004
  .Packet Lenght (2 bytes): x3455
  .Idle Data: x55..55
CLCW: x010004c9
  .Type (1 bit): b0
  .Version Number (2 bits): b00
  .Status (3 bits): b000
  .COP Use (2 bits): b01
  .Virtual Channel ID (6 bits): b000000
  .Reserved (2 bits): b00
  .Flags (5 bits): b00000
  .FARM-B Counter (2 bits): b10
  .Reserved (1 bit): b0
  .Informed Value (1 byte): bc9
FECW (2 bytes): xeb62
   Test Procedure Finished
```

As a result of this command, the CPDU output signal was acquired using the Tektronix MSO-2024B oscilloscope and Figure 68 presents the result in which it is possible to identify a pulse of 13 ms (since the time scale is of 4 ms) and two cursors identify the rising and the falling edge.



Figure 68 – Output 13ms pulse from a CPDU command.

Source: acquired with the Tektronix MSO-2024B.

5.2.4 Negative-acknowledgement Error Codes

The communication module applies a verification process when a direct telecommand (CPDU command) is received by the UTMC, and when the OBDH received a routed telecommand from the UTMC.

This process consists of a data consistency check related to the protocol structure. As a result of this verification, a report packet is generated containing the source command header, and when an error is detected, the packet consists in the annex of an error code with the source command header. Table 11 presents the UTMC possible values of the identification error field.

However, the OBDH only generates the confirmation packet under request or if a CRC error occurs, since it is not possible to define if this error was in the "Ack Flag" field. The packet follow the same pattern of those generated by the UTMC system and the possible values of the "Error Code" field are presented in Table 12.

Error Code	Description
x00	Illegal APID
x01	Incomplete or invalid length packet
x02	Incorrect CRC
x03	Illegal packet type
x05	Illegal or inconsistent application data
x08	Invalid version number
x09	Invalid data field header flag

Table 11 – Direct telecommand NACK error codes.

Source: from (BEZERRA et al., 2010).

Table 12 – OBDH Verification NACK error codes.

Error Code	Description
x001	Incorrect CRC
x002	Invalid version number
x004	Illegal packet type
x008	Illegal or inconsistent application data
x010	Invalid sequence flag
x020	Invalid PUS version number
x040	Invalid ACK flag
x080	Illegal service type
x100	Illegal message subtype

Source: from author.

To test this verification functionality, a CPDU command with a known forced error in the "Version Number" field is sent to the communication module. The expected answer should contain in the "Error Code" field the '08'. Figure 69 presents the CPDU telecommand structure, where the inserted error is written in red in the "Packet Version Number" field. Figures 70, 71 and 72 presents the received report packet.

Figure 69 - CPDU telecommand structure with inseted error.

∢			- •TC	Transfer Fran	ne Pri	imary Hea	der-				>
Version Number 2 bits	Bypass Flag 1 bit	bass ag bit bit Command Flag 1 bit		Reserved Spare 2 bits	Spacecraft Identifier 10 bits		Virtual Channel Identifier 6 bits		Frame Length 10 bits		Frame Sequence Number 8 bits
b00	b0	0b		b00 b11111111		11111111	b	000000 60000010		0001	xC8
∢				Packet Prim	ary H	leader					>
	Packet Identification					Packet Sequence Control			ontrol	Packet	
Packet Version Number 3 bits	Packet Type 1 bit	Second Heade Flag 1 bit	ondary Application eader Process Flag Identifier 1 bit 11 bits		n	Sequen Flags 2 bits	ice 3 3	Packet Sequence Count or Packet Name 14 bits		I	Data Length 6 bits
<i>b010</i>	bl	b0		ь000000000	001	b11		600000000000000000000000000000000000000		3	x0003
∢		User Data Fie	ld		>						
	Source Data										
CPDU Output 12 bit	CPDU Output Number Reserved Register Value 12 bits 1 bit 3 bits		;	CRC FEC 16 bits 16 bits		s					
60000000	00000	b0		b000		x	1D0	F	,	.9C0	7

Figure 70 – CPDU with error command flow: telecommand to be sent.

lucas@cubos-licenses:~/Projects/payload-x/scripts\$ python3 test_procedure.py

```
Payload-X Test Procedure
=> Choose the use case:
  1 - Destination: OBDH Config | Expected: Ack
  2 - Destination: OBDH Status | Expected: Status Report
  3 - Destination: UTMC CPDU | Expected: Ack
  4 - Destination: UTMC CPDU
                             | Expected: NAck (code error x08)
  5 - Destination: OBDH
                              | Expected: NAck (code error x080)
Commando: 4
=> Telecommand (TT&C -> Payload-X):
Unlock Command
Generated CLTU:
 Start Sequence (2 bytes): xEB90
 BCH Code Block (8 bytes): x33FF0007000030AA
 BCH Code Block (8 bytes): x430000000000078
 Stop Sequence (8 bytes): xC5C5C5C5C5C5C579
Set Command
Generated CLTU:
 Start Sequence (2 bytes): xEB90
 BCH Code Block (8 bytes): x33FF000900820052
 BCH Code Block (8 bytes): xC8BFE4000000004A
 Stop Sequence (8 bytes): xC5C5C5C5C5C5C579
CPDU Command
Generated CLTU:
 Start Sequence (2 bytes): xEB90
 BCH Code Block (8 bytes): x03FF0011C8C0501E
 BCH Code Block (8 bytes): x01C000000300002E
 BCH Code Block (8 bytes): x1D0FEE1D000000FE
 Stop Sequence (8 bytes): xC5C5C5C5C5C5C579
       Source: from author.
```

In Figure 71 the value '01' in the "Service Type ID" identifies that a confirmation message and the value '02' in the "Message Type ID" indicates that the message is a NACK report. The last byte of the "User Data" field is the error code from the CPDU unit.

Figure 71 – CPDU with error command flow: expected telemetry (first part).

```
=> Telemetry (Payload-X -> TT&C):
ASM: x1acffc1d
Transfer Frame Primary Header: x3ff300001800
  .Version Number (2 bits): b00
  .Virtual Channel ID (3 bits): b001
  .Operational Control Field Flag (1 bit): b1
  .Master Channel Frame Counter (1 byte): x00
  .Virtual Channel Frame Counter (1 byte): x00
  .Transfer Frame Data Field Status (2 bytes): x1800
Space Packet Primary Header: x0801c0000010
  .Packet Version Number (3 bits): b000
  .Packet Type (1 bit): b0
  .Application Process ID (11 bits): b0000000001
  .Sequence Flag (2 bits): b11
  .Sequence Count (14 bits): xc000
  .Packet Lenght (2 bytes): x0010
Space Packet Secondary Header: x200102000000000
  .TM PUS Version Number (4 bits): b0010
  .Spacecraft Time (4 bits): b0000
  .Service Type ID (1 byte): x01
  .Message Type ID (1 byte): x02
  .Message Type Counter (2 byte): x0000
  .Destination ID (2 bytes): x0000
  .Time (1 byte): x00
User Data : x5001c00008
CRC (2 bytes): xd742
```

Figure 72 – CPDU with error command flow: expected telemetry (second part).

```
Idle Packet: x07ffc000043355..55
  .Packet Version Number (3 bits): b000
  .Packet Type (1 bit): b0
  .Secondary Header Flag (1 bit): b0
  .Application Process ID (11 bits): b1111111111
  .Sequence Flag (2 bits): b11
  .Sequence Count (14 bits): xc000
  .Packet Lenght (2 bytes): x0433
  .Idle Data: x55..55
CLCW: x010004c9
  .Type (1 bit): b0
  .Version Number (2 bits): b00
  .Status (3 bits): b000
  .COP Use (2 bits): b01
  .Virtual Channel ID (6 bits): b000000
  .Reserved (2 bits): b00
  .Flags (5 bits): b00000
  .FARM-B Counter (2 bits): b10
  .Reserved (1 bit): b0
  .Informed Value (1 byte): bc9
FECW (2 bytes): xa033
      Test Procedure Finished
     Source: from author.
```

The second case, a telecomand using a wrong destination for the OBDH system is sent using the test setup. Figure 73 presents the structure of the telecommand. Figures 74, 75 and 76 presents the received report packet.

Figure 73 – OBDH command structure with wrong destination.

		TC	C Transfe	er Frame	e Pri	mary Head	der-					
Version Number 2 bits	Bypass Flag 1 bit	Control Command Flag 1 bit	Reser Spa 2 bi	rved are its	Sp Id	acecraft entifier 10 bits	C Id	/irtual hannel entifier 6 bits	Frame Lengt 10 bit	e h s	Fram Sequer Numb 8 bit	
b00	b0	0b	b0	b00 b111111111 b		b	000001	b0000010100		x64		
			Packe	et Primar	ry H	eader					>	
		Packet Identification			Pack	Packet Sequence Control			Paakat			
Packet Version Number 3 bits	Packet Type 1 bit	Secondary Header Flag 1 bit	Application Process Identifier 11 bits		l	Sequen Flags 2 bits	Sequence Cou Flags Packet S 2 bits 14		Sequence nt or I t Name bits		Data Length 16 bits	
ь000	b1	b1	b1 b0000000000		b11	ь000000000000000		3	x0006			
		TC Pac	ket Seco	ondary H	Head	ler		·····>	,			
		Ν	Лessage	Type ID)							
TC PUS Version Number	Ack Flags 4 bits	Service Typ 8 bits	e ID Message Su 8 bi		ibtype ID ts	Sc	ource ID 16 bits					
4 0113		x06						1				

CRC	FEC
16 bits	16 bits
xB99C	xCAD1

Figure 74 – Command flow of a OBDH command with wrong destination: telecommand to be sent.

lucas@cubos-licenses:~/Projects/payload-x/scripts\$ python3 test_procedure.py

Payload-X Test Procedure => Choose the use case: 1 - Destination: OBDH Config | Expected: Ack 2 - Destination: OBDH Status | Expected: Status Report 3 - Destination: UTMC CPDU | Expected: Ack 4 - Destination: UTMC CPDU | Expected: NAck (code error x08) 5 - Destination: OBDH | Expected: NAck (code error x080) Commando: 5 => Telecommand (TT&C -> Payload-X): Unlock Command Generated CLTU: Start Sequence (2 bytes): xEB90 BCH Code Block (8 bytes): x33FF04070000FA52 BCH Code Block (8 bytes): xB2000000000008A Stop Sequence (8 bytes): xC5C5C5C5C5C5C579 ___Set Command___ Generated CLTU: Start Sequence (2 bytes): xEB90 BCH Code Block (8 bytes): x33FF0409008200EE BCH Code Block (8 bytes): x64CD23000000014 Stop Sequence (8 bytes): xC5C5C5C5C5C5C579 OBDH Command Generated CLTU: Start Sequence (2 bytes): xEB90 BCH Code Block (8 bytes): x03FF041464C0182A BCH Code Block (8 bytes): x00C00000628062A BCH Code Block (8 bytes): x070000B99CCAD18C Stop Sequence (8 bytes): xC5C5C5C5C5C5C579 Source: from author.

In Figure 75 the value '01' in the "Service Type ID" indicates a confirmation message, while the value '02' in the "Message Type ID" indicates that the message is a NACK report. The 2 first bytes in the "User Data" are the error code and the last four bytes indicates the Space Packet header of the source telecommand.

Figure 75 – Command flow of a OBDH command with wrong destination: expected telemetry (first part).

```
=> Telemetry (Payload-X -> TT&C):
ASM: x1acffc1d
Transfer Frame Primary Header: x3ff304041800
  .Version Number (2 bits): b00
  .Virtual Channel ID (3 bits): b001
  .Operational Control Field Flag (1 bit): b1
  .Master Channel Frame Counter (1 byte): x04
  .Virtual Channel Frame Counter (1 byte): x04
  .Transfer Frame Data Field Status (2 bytes): x1800
Space Packet Primary Header: x0800c0040010
  .Packet Version Number (3 bits): b000
  .Packet Type (1 bit): b0
  .Application Process ID (11 bits): b0000000000
  .Sequence Flag (2 bits): b11
  .Sequence Count (14 bits): xc004
  .Packet Lenght (2 bytes): x0010
Space Packet Secondary Header: x20010200040000000000000
  .TM PUS Version Number (4 bits): b0010
  .Spacecraft Time (4 bits): b0000
  .Service Type ID (1 byte): x01
  .Message Type ID (1 byte): x02
  .Message Type Counter (2 byte): x0004
  .Destination ID (2 bytes): x0000
  .Time (1 byte): x00
  .Spare (1 byte): x00
User Data : x00801800c000
CRC (2 bytes): xd9d7
```

Figure 76 – Command flow of a OBDH command with wrong destination: expected telemetry (second part).

```
Idle Packet: x07ffc000043155..55
  .Packet Version Number (3 bits): b000
  .Packet Type (1 bit): b0
  .Secondary Header Flag (1 bit): b0
  .Application Process ID (11 bits): b1111111111
  .Sequence Flag (2 bits): b11
  .Sequence Count (14 bits): xc000
  .Packet Lenght (2 bytes): x0431
  .Idle Data: x55..55
CLCW: x01040465
  .Type (1 bit): b0
  .Version Number (2 bits): b00
  .Status (3 bits): b000
  .COP Use (2 bits): b01
  .Virtual Channel ID (6 bits): b000001
  .Reserved (2 bits): b00
  .Flags (5 bits): b00000
  .FARM-B Counter (2 bits): b10
  .Reserved (1 bit): b0
  .Informed Value (1 byte): b65
FECW (2 bytes): xbc9c
        Test Procedure Finished
```

5.3 CORTEX RESULTS

During the development of this work, it was possible to have access to Cortex equipment. However, the available time was limited, and the OBDH system was not ready at that moment. Therefore the test focused on a continuous transmission in the telemetry flow using the Synchronous Serial Interface (SSI) as communication interface, with a data rate of 500 kbps. The Cortex equipment is capable of verifying the consistency of the data and performs the decoding using the Reed-Solomon algorithm.

Considering the information presented above, the embedded application in the development kit was only the UTMC, and it was configured for a continuous transmission mode of idle packets. Two different approaches were tested. For the first one, the telemetry frame was uncoded and, in the second one, the Reed-Solomon algorithm was applied.

Fig. 77 shows a received idle packet consisting of several 55_{16} (55 in hexadecimal) bytes using the embedded software provided by the equipment. The counter is at 3466 showing the compatibility of the packets sent by the UTMC, since wrong data would reset the counter.

			Levels Level and Level 201
IFR Unlocke	d PSK Un	locked B/S Un	locked F/S Locked
Frame		Display	1
Time 1-	01:24:57:534	Raw Byte (•	Frame Sampling 36 🕂
Counter	3466	Baw Short C	
Check	Ok	Ban Long C	insert Decom.
C	- OK	naw Long	Disable
Sync.Status	Locked	Decom. L3	
▲ Offset	Telemetry Frames		
0180	55 55 55 55 55 55	55 55 55 55 55 55 55 5	5 55 55 55 55 55 55 55 55 55 55 5
01C0	55 55 55 55 55 55	55 55 55 55 55 55 55 55	5 55 55 55 55 55 55 55 55 55 55 55
0200	55 55 55 55 55 55	55 55 55 55 55 55 55 5	5 55 55 55 55 55 55 55 55 55 55 55
0240	55 55 55 55 55 55	55 55 55 55 55 55 55 55 5	5 55 55 55 55 55 55 55 55 55 55 55
0280	55 55 55 55 55 55	55 55 55 55 55 55 55 55 5	5 55 55 55 55 55 55 55 55 55 55 55 55
02C0	55 55 55 55 55 55	55 55 55 55 55 55 55 55 5	5 55 55 55 55 55 55 55 55 55 55 55 55 5
0300	55 55 55 55 55 55	55 55 55 55 55 55 55 55	5 55 55 55 55 55 55 55 55 55 55 55 55 5
0340	55 55 55 55 55 55	55 55 55 55 55 55 55 55	5 55 55 55 55 55 55 55 55 55 55 55 55 5
0380	55 55 55 55 55 55	55 55 55 55 55 55 55 55	5 55 55 55 55 55 55 55 55 55 55 55 55 5
03C0	55 55 55 55 55 55	55 55 55 55 55 55 55 55	5 55 55 55 55 55 55 55 55 55 55 55 55 5
0400	55 55 55 55 55 55	55 55 55 55 55 55 55 55	5 55 55 55 55 55 55 55 55 55 55 55 55 5
- 0440	55 55 55 55 55 55	55 55 55 55 55 55 55 55 55	5 55 55 55 55 55 55 55 55 55 55 55 55 5
0110	33 33 33 33 33 33	55 55 55 55 55 55 55 5	0 00 00 00 00 00 00 00 00 00 00 00

Figure 77 – Packet received in CORTEX.

Source: from Cortex equipment.

The logs of the performed tests were saved using the TCP/IP interface provided by the equipment. To represent the acquired results, Table 13 presents the report for un-coded telemetry, while Table 14 presents the telemetry by using Reed-Solomon coding. In both cases, the test was performed several times, and it was stopped to start another one always when the Sequence Counter was higher than 100,000 frames. This means that the implementation was able to send the correct pattern without errors, at least more than 100,000 times.

Table 13 – Cortex report of the un-coded telemetry frame.

Cortex Report - Un-coded Telemetry Frame				
Tolomotry Framo	$1 a c f f c 1 d 3 f f 3 9 5 9 5 1 f f e 5 5 \dots 5 5 0 1 0 0 2 0 0 0$			
relementy frame	a8ab			
Cortex Local Time	0:31:37.000431			
Sequence Counter	103,828			
Frame Check and RS Status	[2] CRC verification ON			
Frame Check Result	[0] Frame Check OK			
Frame Sync Status	[1] Frame sync. ON, LOCK			
Bit Slip Status	0 No bit flip			
TM Delay	0			
Frame Length (bytes)	1119			
Sync Word Length (bytes)	32			

Source: based on the results from Cortex.

Cortex Report	
	1acffc1d3ff367671ffe555501002000
	040 d7 be 5 a 4 f 264 d 43788 b 929 e 1 b f e 233
	cff131b49a01 79 $c67973749099038426$
	12a8560dc34e43291232ae8f32c20996
	ca 287 ca 3a 0 c 0 3271 6319 a 53 a d 4a 72500
Telemetry Frame	2 cfd cace 0 af 024 ff 265 e 3f 8f e f a 5591 d 084
	12 c 4375 c 5 c b 65 c a a 276 e 86 c 7 e f 9287369
	4b927c10f5199cdf986b101440e085772
	edeb2204eba81368c04d388a93db9f87f
	3afd 18f 3f 9ff 4d 284866 ed 415391409 c 5d
	ab2f1e7f9fedc89ef4253b272305
Corte Local Time	1:05:58.000104
Sequence Counter	105,257
Frame Check and RS Status	[3] RS decoder; CRC verification ON
Frame Check Result	[0] Frame Check OK
Frame Sync Status	[1] Frame sync. ON, LOCK
Bit Slip Status	[0] No bit flip
TM Delay	0
Frame Lenght (bytes)	1279
Sync Word Lenght (bytes)	32

Table 14 – Cortex report of the Reed-Solomon telemetry frame.

Source: based on the results from Cortex.

5.4 SYNTHESIS RESULTS

Taking into account the features available in the Brave FPGA, Table 15 presents the results obtained from the NanoXplore synthesis tool. The reference version for the presented results uses BCH coding in the telecommand stream, and the telemetry is generated without codes for error detection and correction.

Moreover, a Static Time Analysis was performed to check if the implementation is capable of running at required frequency. Table 16 presents the results. The required frequency for this design is 25 MHz, and the implementation could run using a maximum frequency of 34.301 MHz. It is important to underline that there is no negative slack.

Instance	Count	Total Available	Usage (%)
4-LUT	7861	34272	23
DFF	3599	32256	12
XLUT	265	2016	14
Carry	2077	8064	26
Register File Block	0	168	0
Clock Switch	0	336	0
DSP Block	0	112	0
Memory Block	5	56	9
WFG	1	32	4
PLL	0	4	0

Table 15 – Synthesis result: resource usage.

Source: based on the results from NanoXplore Synthesis tool.

Static Time Analisys Report					
Source	Target	Hold/Remo	val Summary		
		Slack	Min. Data		
		Sluck	Arrive Time		
Input	Clock (falling)	-	1.095 ns		
Input	Clock (rising)	-	3.022 ns		
Clock (falling)	Clock (falling)	1.248 ns	1.248 ns		
Clock (falling)	Clock (rising)	$21.250 \mathrm{ns}$	$1.250 \mathrm{ns}$		
Clock (rising)	Clock (falling)	$21.250 \mathrm{ns}$	$1.250 \mathrm{ns}$		
Clock (rising)	Clock (rising)	1.243 ns	$1.243 \mathrm{ns}$		
Clock (rising)	Output	-	$7.923 \mathrm{ns}$		
Source	Target	Setup/Record	ery Summary		
		Slack	Min. Data		
		Sluck	Arrive Time		
Input	Clock (falling)	-	$6.643 \mathrm{ns}$		
Input	Clock (rising)	-	$10.097 \mathrm{ns}$		
Clock (falling)	Clock (falling)	$28.753 \mathrm{ns}$	11.247 ns		
Clock (falling)	Clock (rising)	5.423 ns	14.577 ns		
Clock (rising)	Clock (falling)	$6.320 \mathrm{ns}$	$13.680 \mathrm{ns}$		
Clock (rising)	Clock (rising)	23.302 ns	$16.698 \mathrm{ns}$		
Clock (rising)	Output	-	15.896 ns		
Frequ	ıency	Required	Maximum		
		$25.000 \mathrm{~MHz}$	34.301 MHz		

Table 16 – Static Time Analisys results.

Source: based on the results from NanoXplore Synthesis tool.

6 CONCLUSION

This work presented the development of a communication module for nanosatellites based on a hardware application. The main focus of this work was the construction of an architecture based on the CCSDS recommendations. The implemented architecture is composed of a Unit of Telemetry and Telecommands (UTMC) and an OBDH (On-Board Data Handling). The UTMC works in the scope of the Data Link Layer described in the CCSDS 130.0-G-3 - Overview of Space Communications Protocols, since this unit handles the data structure presented by the recommendations that embrace this layer and performs the data coding and decoding. The developed OBDH performs the Network Layer and uses the Space Packet Protocol. The OBDH was implemented to be able to validate the received telecommands and generate status and scientific data to be sent within the telemetry framework.

The proposed architecture represents a trade off between meeting the specific demands of the case study and being generic enough to be applied in other scenarios. The implementation was first simulated through the tool ModelSim and next it was physically tested with the help of the NanoXplore development board.

It is worth noting that after the implementation of the physical synthesis, it was verified that the FPGA could support not only the current implementation, but also additional control modules. Thereby, aiming at the development of a complete control system for a CubeSat mission within an FPGA, a future expansion of the system will be achieved possible thank to more capable FPGAs that the company NanoXplore plans to produce.

Since the presented results are satisfactory for the proposed application, it is possible to conclude that the work achieved the intended objectives. It is important to note that the work has continuity in the scope of the FloripaSat-I mission since it is embedded in the developed architecture for Payload-X, which is the result of team work within the research group.

6.1 ACADEMIC PRODUCTION

In the framework of this project, two research articles were submitted and accepted as regular papers for oral presentation in international conference of the domain. 1. LUZA, L. M.; RIGO, C. A.; TRAMONTIN, E. D.; MARTINS, V.; MARTINEZ, S. V; SLONGO, L. K.; SEMAN, L. O.; DILILLO, L.; BEZERRA, E. A. Enabling deep-space CubeSat missions through state-of-the-art radiation-hardned technologies. In: *III IAA Latin American CubeSat Workshop (LACW-IAA 2018)*, Ubatuba, São Paulo, 2018.

2. RIGO, C. A.; LUZA, L. M.; TRAMONTIN, E. D.; MARTINS, V.; MARTINEZ, S. V; SLONGO, L. K.; SEMAN, L. O.; DILILLO, L.; VARGAS, F. L.; BEZERRA, E. A. A Fault-Tolerant Reconfigurable Platform for Communication Modules of Satellites *The 20th IEEE Latin-American Test Symposium (LATS 2019)*, Santiago, Chile, 2019.

6.2 FUTURE WORKS

During the development of this work the following possibilities of future work have been identified:

- Validation of the system using the Payload-X architecture. A development kit was used to test the proposed communication module, since the access to a Payload-X board was yet not possible until the end of this work. Thus, there is still a need to test the final implementation with all the components interconnected.
- Implementation of the SRAM-based controller monitor. Once the memory controller will be available by a partner of FloripaSat project, the integration with the OBDH described in this work must be performed;
- PUS Services implementation. The proposed work used the ECSS PUS frame structure to define custom services using the OBDH system. For future works, it would be interesting the addition of the default services proposed by the recommendation since it increases the possibility to reuse the architecture;
- Addition of an on-chip bus. The actual work can use two different protocols to do the communication between the UTMC and OBDH, the Synchronous Serial Interface (SSI) and Universal Asynchronous Receiver/Transmitter (UART). However, an onchip interconnection bus as Advanced Microcontroller Bus Archi-

tecture (AMBA) could provide a higher frequency communication between the entities.

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FSM With Three Processes VHDL Coding Example.

```
library IEEE;
use IEEE.std_logic_1164.all;
entity fsm_3 is
 port (clk, reset, input : in std_logic;
       output
                            : out std_logic
 );
end entity;
architecture behavioral of fsm_3 is
   type state_type is (s1,s2,s3,s4);
   signal state, next_state: state_type;
begin
   process1 : process (clk, reset)
       begin
           if (reset ='1') then
               state <= s1:
           elsif (clk='1' and clk'Event) then
               state <= next_state;</pre>
           end if;
   end process process1;
   process2 : process (state, input)
       begin
       case state is
           when s1 =>
               if (input = '1') then
                   next_state <= s2;</pre>
               else
                   next_state <= s3;</pre>
               end if;
           when s2 =>
               next_state <= s4;</pre>
           when s3 =>
               next_state <= s4;</pre>
           when s4 =>
               next_state <= s1;</pre>
       end case;
   end process process2;
   process3 : process (state)
       begin
       case state is
```

```
when s1 =>
        output <= '1';
        when s2 =>
            output <= '1';
        when s3 =>
            output <= '0';
        when s4 =>
            output <= '0';
    end case;
end process process3;</pre>
```

end behavioral;

Source: (XILINX; INC, 2002).